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SIMULATION OF A STANDING-WAVE FREE-ELECTRON LASER*

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Abstract

The standing-wave free-electron laser (FEL) differs from a conventional linear-wiggler microwave FEL in using irises along the wiggler to form a series of standing-wave cavities and in reaccelerating the beam between cavities to maintain the average energy. The device has been proposed for use in a two-beam accelerator (TBA) because microwave power can be extracted more effectively than from a traveling-wave FEL. The standing-wave FEL is modeled in the continuum limit by a set of equations describing the coupling of a one-dimensional beam to a TE₀₁ rectangular-waveguide mode. Analytic calculations and numerical simulations are used to determine the time variation of the reacceleration field and the prebunching required so that the final microwave energy is the same in all cavities. The microwave energy and phase are found to be insensitive to modest spreads in the beam energy and phase and to errors in the reacceleration field and the beam current, but the output phase appears sensitive to beam-energy errors and to timing jitter.

Introduction

The next generation of linear colliders is expected to require accelerating gradients of 100 MeV/m or greater. For the high-gradient structures that have been tested, this field strength corresponds to a microwave power of about 100 MW/m, and the required frequency is typically in the range of 10-30 GHz. The microwave free-electron laser (FEL)¹ and the relativistic klystron (RK)² have both demonstrated the required power level in this frequency range, and they have been proposed as collider power sources in a configuration known as the "two-beam accelerator" (TBA)³, in which a high-current "drive" beam generates microwave energy in a beamline that parallels the high-gradient structure. Both the RK/TBA and the FEL/TBA have practical problems. The RK operates best in the X-band (8.4-12 GHz) and so cannot be used with many high-gradient accelerator designs. Conventional FELs, in contrast, have no fundamental frequency limitation, but experiments have shown that microwave extraction is difficult.⁴

The cavity-coupled FEL/TBA⁵ has been devised to sidestep the problems found with RKs and conventional FELs. This new device would replace the usual FEL waveguide by a series of short standing-wave cavities, each about a wiggler period in length. The cavities would be separated by irises that allow the beam to pass but reflect most of the microwave power, and between cavities there would be induction accelerating cells to

maintain the beam energy. Microwave energy from these cavities would be connected to the high-gradient structure by couplers and would oscillate between the two beam lines with a period that is much longer than the beam time scale but much less than the resistive loss time. This coupling scheme was proposed by Henke⁶ for a RK/TBA and is discussed elsewhere.⁵

In this paper, we present preliminary numerical simulations of the standing-wave FEL (SWFEL) used in the cavity-coupled FEL/TBA. The SWFEL has two important differences from conventional FEL amplifiers. One difference is that the standing-wave phase ϕ develops in time only at each cavity location z , whereas the wave phase in conventional devices evolves in z along with the "particle phase" $\theta_j = (k_s + k_w)z - \omega_s t$, where the subscript j denotes the j th particle. This difference works against the preservation of a nearly constant average bucket phase $\langle \psi_j \rangle = \langle \theta_j + \phi \rangle$ that is needed for good bunching. As a consequence, the SWFEL requires an unusual form of prebunching, as we discuss later. A second difference is the use of frequent reacceleration to maintain a nearly constant beam energy. Reacceleration is used rather than tapering of the wiggler strength because it is more appropriate for the very long beamlines expected in linear colliders and because it in principle allows the beam energy to be adjusted in time as well as in z . Since the unusual phase evolution is a critical novel aspect of SWFEL physics, we choose a very simple simulation model that retains this feature but ignores other arguably important features, such as the discrete nature of the standing-wave cavities and the competition between waveguide modes. The next section describes this model briefly and is followed by a section on simulation results. We offer some tentative conclusions in a final section.

Model

Assumptions and Equations

Simulation particles are modeled by a pair of wiggle-averaged equations for the total energy γ_j in units of $m_e c^2$ and the particle phase θ_j . Radial motion and the effects of the transverse beam structure are neglected, and the beam is assumed to couple only with a TE₀₁ waveguide mode, which is usually most strongly coupled mode. The signal wavenumber for this mode in a rectangular waveguide with height h and width w is $k_s = (\omega_s^2/c^2 - \pi^2/h^2)^{1/2}$. For the fields, we assume an idealized linear wiggler with a vector potential

$$\tilde{A}_w = \frac{m_e c^2}{e} a_w \cos(k_w z) \hat{x} \quad (1)$$

and an appropriate form for the signal field

$$\tilde{A}_s = -\frac{m_e c^2}{e} a_s q_\perp(x, y) \cos(k_s z - \omega_s t + \phi), \quad (2)$$

where $q_\perp = -\sin(\pi y/h) \hat{x}$ is the transverse structure for a TE₀₁ mode. A number of other conventional assumptions are made that are suitable for most Compton-regime FELs and significantly simplify the equations. The energy is taken to be sufficiently high that $a_w/\gamma_j \ll 1$, and the energy spread is assumed small enough that all particles have effectively the same axial velocity V_b . We treat the signal amplitude a_s as small

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compared with a_w , and the both a_s and ϕ are assumed to be slowly varying compared with k_z and $\omega_w t$. This last assumption makes the equations inappropriate for modeling waveguide modes near cutoff.

The wiggler-averaged particle equations are identical to those in a conventional single-mode microwave FEL. Taking z to be the independent variable, we write the equations as

$$\frac{d\theta_j}{dz} = k_w + k_s - \frac{\omega_s}{c} - \frac{\omega_s}{2c\gamma_j^2} \left[1 + \frac{a_w^2}{2} - 2D_x a_w (\hat{a}_r \cos \theta_j - \hat{a}_i \sin \theta_j) \right] \quad (3a)$$

$$\frac{d\gamma_j}{dz} = -D_x \frac{\omega_s}{c} \frac{a_w}{\gamma_j} (\hat{a}_r \sin \theta_j + \hat{a}_i \cos \theta_j) - \frac{eE_z}{m_e c^2}. \quad (3b)$$

Here, the coupling coefficient D_x is given for a TE₀₁ mode by

$$D_x = \frac{1}{2} [J_0(\xi) - J_1(\xi)], \quad (4)$$

where $\xi = \omega_s a_w^2 / (8ck_w \gamma_j^2) \approx a_w^2 / 4(1 + a_w^2/2)$. An equation for the complex signal amplitude $\hat{a} \equiv \hat{a}_r + i\hat{a}_i = a_s \exp(i\phi)$ is obtained by assuming that \hat{a} evolves only in time and requiring that the wiggler-averaged equations conserve energy. Taking the distance back from the beam head $s \equiv V_b z - t$ as the "time" coordinate, this procedure gives the field equation

$$\frac{\partial \hat{a}}{\partial s} = i\eta \left\langle \frac{\exp(-i\theta_j)}{\gamma_j} \right\rangle, \quad (5)$$

where the coefficient η in general depends on s and is given by

$$\eta = \frac{4\pi}{hw} \frac{eI_b}{m_e c^3} \frac{c}{V_b} \frac{c}{\omega_s} D_x a_w. \quad (6)$$

While this equation implicitly assumes an infinitesimal cavity length and ignores field coupling through the cavity irises, it does model the novel signal evolution expected in a SWFEL.

"Single-Particle" Solution

Some understanding of the SWFEL equations is gained by looking at a z -independent "single-particle" solution, in which the full beam current is assigned to a single phase-space point. Linearizing the equations for small $\delta\gamma = \gamma - \gamma_r$, where $\gamma_r^2 = \omega_s(1 + a_w^2/2)/2c(k_w + k_s - \omega_s/c)$ is the resonant energy, we obtain the approximate particle equations

$$\frac{d\theta}{dz} \approx 2 \left(k_w + k_s - \frac{\omega_s}{c} \right) \frac{\delta\gamma}{\gamma_r} \quad (7a)$$

$$\frac{d\delta\gamma}{dz} \approx -D_x \frac{\omega_s}{c} \frac{a_w}{\gamma_r} (\hat{a}_r \sin \theta + \hat{a}_i \cos \theta) - \frac{eE_z}{m_e c^2}. \quad (7b)$$

Requiring $\delta\gamma$ to be z -independent gives the E_z required for equilibrium:

$$\frac{eE_z}{m_e c^2} \approx -D_x \frac{\omega_s}{c} \frac{a_w}{\gamma_r} (\hat{a}_r \sin \theta + \hat{a}_i \cos \theta). \quad (8)$$

If $\delta\gamma$ is initially zero, then θ is likewise independent of z and equal to some arbitrary $\theta_0(s)$. The components of \hat{a} in Eq. (8) are obtained by integrating the linearized field equation, which gives

$$\hat{a}(s) = \hat{a}(0) + \frac{i}{\gamma_r} \int_0^s ds' \eta(s') \exp[-i\theta_0(s')]. \quad (9)$$

As a practical special case, we consider a beam with constant η which is prebunched at a frequency $\omega_s + \Delta\omega$, so that $\theta_0(s) = \alpha - (\Delta\omega/V_b)s \equiv \alpha + \beta s$. The components of \hat{a} are then given by

$$\hat{a}_r(s) = \hat{a}_r(0) + \frac{\eta}{\beta\gamma_r} [\cos(\alpha) - \cos(\alpha + \beta s)] \quad (10a)$$

$$\hat{a}_i(s) = \hat{a}_i(0) - \frac{\eta}{\beta\gamma_r} [\sin(\alpha) - \sin(\alpha + \beta s)], \quad (10b)$$

and the corresponding reacceleration field is

$$\frac{eE_z}{m_e c^2} = -D_x \frac{\omega_s}{c} \frac{a_w}{\gamma_r} \left[\hat{a}_r(0) \sin(\alpha + \beta s) + \hat{a}_i(0) \cos(\alpha + \beta s) + \frac{\eta}{\beta\gamma_r} \sin(\beta s) \right]. \quad (11)$$

The bucket size shrinks with increasing $E_z/|\hat{a}|$, and for this case it is straightforward to calculate the minimum extent of the bucket in θ and γ as the signal develops in s . These minimum values are approximately

$$\Delta\theta \approx 3.4 \left(-\frac{\beta\gamma_r |\hat{a}(0)|}{\eta} \right)^{1/2} \quad (12a)$$

$$\Delta\gamma \approx 3.3 \left(\frac{\omega_s}{c} \frac{D_x a_w}{k_w - \delta k} \right)^{1/2} \left(-\frac{\beta\gamma_r}{\eta} \right)^{1/2} |\hat{a}(0)|, \quad (12b)$$

where $\delta k = \omega_s/c - k_s$. This result shows that the bucket vanishes when β is zero or negative and that the longitudinal acceptance $\Delta\theta$ $\Delta\gamma$ increases with a larger initial signal and larger $-\beta/\eta$. It is found that Eq. (12) underestimates the acceptance for distributions with spreads in θ and γ because the required reacceleration field in such cases is somewhat lower than for the single-particle case. There is also a weak dependence of the acceptance on the initial average particle phase α , with the largest acceptance occurring for $\alpha + \phi(0) = 0$.

Simulation

Parameters

The operating frequency ω_s and the final energy per unit length W_{out} left in the cavities are determined in practice by the TBA requirements. With these quantities given, the specification of the waveguide dimensions h and w , the wiggler wavelength $\lambda_w = 2\pi/k_w$, and the wiggler strength a_w sets the principle beam parameters. The beam energy is determined by the resonance condition, and the total beam charge, given by $I_b L_b$ when the current is constant, is set by W_{out} . Since the initial spreads in θ and γ are usually determined by the intrinsic emittance from the accelerator and the additional emittance introduced by prebunching, the values are not considered free parameters.

Two remaining beam quantities, the beam-current envelope $I_b(s)/\max I_b$ and the prebunching factor β can be chosen by practical considerations. Since the acceptance is found to be proportional to I_b^{-1} , it is preferable for the current to be low near the beam head, where the bucket reaches its minimum size. It is also found from the single-particle equations that a current that increases like s or faster leads to a monotonically increasing E_z for $s \leq L_b$, which is an easier field to generate than a short pulse. For these reasons, we use a beam with a uniform current ramp as our standard case. The prebunching factor is chosen by considering the β -dependences of various beam quantities in the single-particle solution. We find that the

Table 1 Nominal standing-wave FEL parameters

peak beam current	I_b	2.17 kA
beam length	L_b	180.0 cm
initial energy	γ_r	27.6
initial θ -spread	$\Delta\theta_0/2\pi$	0.1
initial γ -spread	$\Delta\gamma_0/\gamma_r$	0.01
wiggler strength	a_w	8.86
wiggler wavelength	λ_w	25 cm
wiggler length	L_w	40 m
waveguide height	h	3 cm
waveguide width	w	10 cm
signal frequency	$\omega_s/2\pi$	17.1 GHz
cavity Q	Q	10^4
input power	P_{in}	5 kW/m
output energy	W_{out}	10 J/m

required beam charge and the longitudinal acceptance increase with βL_b , while the maximum reacceleration field decreases. Since the beam emittance is difficult to decrease in induction accelerators, we choose $\beta L_b = \pi$, although a lower value might be selected if the limited acceptance of the SWFEL is not found to be a problem.

The nominal parameters used in the simulations here are listed in Table 1. These values are appropriate for a generic TBA, and little effort has been made to optimize the waveguide size or the wiggler strength and wavelength.

Initialization

The simulation initialization parallels the single-particle solution. A distribution with prescribed spreads $\Delta\theta_0$ and $\Delta\gamma_0$ in θ_j and γ_j is loaded so that $\langle\theta_j\rangle = \alpha + \beta s$ and $\langle\gamma_j\rangle = \gamma_r$. Simulation particles are uniformly distributed within this phase-space rectangle, and different random position are chosen for each beam slice. Such a distribution is not realistic, but it allows the longitudinal acceptance to be tested systematically. For the small spreads in θ_j and γ_j treated here, 200 simulation particles are adequate to give acceptably low statistical noise.

The reacceleration field required to keep $\langle\gamma_j\rangle$ constant is given by

$$\frac{eE_z}{m_e c^2} = -D_x \frac{\omega_s}{c} a_w \left(\hat{a}_r \left\langle \frac{\sin \theta_j}{\gamma_j} \right\rangle + \hat{a}_i \left\langle \frac{\cos \theta_j}{\gamma_j} \right\rangle \right). \quad (13)$$

This field could be recalculated at each z and s value, but this algorithm introduces a high-frequency noise component in E_z that increases exponentially with z . A more practical approach is to calculate $E_z(s)$ at $z = 0$ and to use it at all subsequent z positions. With this second technique, the calculated E_z is noise free and reduces to Eq. (11) in the limit that $\Delta\theta_0$ and $\Delta\gamma_0$ are zero.

We set the initial signal level $|\hat{a}(0)|$ by assuming some input microwave power per unit length P_{in} and balancing this with cavity-wall losses, specified by an assumed cavity Q .

Results

The output microwave energy W_{out} and phase ϕ for a beam with the nominal parameters and a linearly increasing I_b are shown in Fig. 1. The spreads $\Delta\theta_0/2\pi = 0.1$ and $\Delta\gamma_0/\gamma_r = 0.01$ used here are small enough that the distribution remains trapped and the output signal is reasonably insensitive to beam and field errors. The principle z dependence in this case is the initial ripple in W_{out} due to synchrotron motion, which corresponds to a 2% fluctuation in the average electron energy. This ripple does not damp fully in the 40 m wiggler because

a deeply trapped distribution randomizes very slowly. There is also a low-amplitude ripple in the wave phase ϕ that results from fluctuations in $\langle \cos(\theta_j + \phi) / \gamma_j \rangle$, due again to synchrotron motion. The wavelength in z of this phase ripple corresponds to the synchrotron wavelength in the initial field because, according to Eq. (5), $\partial\phi/\partial s$ is proportional to $|\hat{a}|^{-1}$, which is largest at small s .

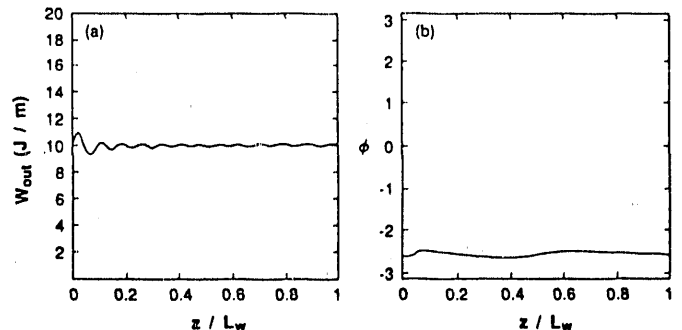


Fig. 1 Output energy per unit length W_{out} and wave phase ϕ as functions of z for the standard case.

For the standard case, the greatest sensitivity to parameter errors is found for fluctuations in the initial energy. When the reacceleration field is calculated for a beam at the resonant energy and the simulation is run with an energy that is 1% higher, W_{out} is nearly unaffected, but ϕ , shown in Fig. 2a, has a ripple of about $\pi/2$. As in the case with no detuning, the ripple wavelength corresponds to the synchrotron wavelength in the initial field, but the amplitude is significantly larger because the distribution centroid is well away from the bucket center and executes large orbits in θ . This phase ripple can be reduced by choosing a larger $|\hat{a}(0)|$, which makes the initial bucket larger, or by decreasing a_w while adjusting λ_w or the waveguide dimensions to maintain a constant resonant energy. Phase ripple is also introduced by variations in the average energy with s , which can develop in an accelerator due to beam loading. As an illustration, Fig. 2b shows ϕ for a beam with an energy equaling γ_r at the beam head but dropping gradually by 4% toward the beam tail. The phase ripple for this case is similar to the constant-energy case in Fig. 1 because the beam distribution remains near the bucket center while the signal amplitude is small.

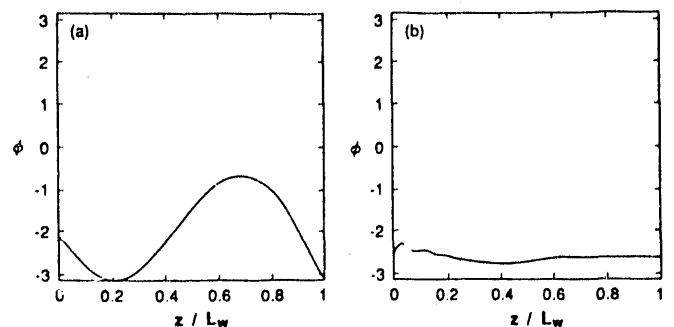


Fig. 2 Wave phase ϕ as a function of z for a beam with (a) a constant energy 1% above γ_r and (b) an energy that decreases by 4% toward the tail.

In contrast to the sensitivity to detuning, a 2% error in I_b has a negligible effect on either W_{out} or ϕ . A change of 2% in the magnitude of E_z likewise has little effect on either the output energy or phase for the standard case, but introducing a 0.1 ns time lag in the reacceleration field again causes a long-

wavelength ripple of about $\pi/2$ in ϕ , as shown in Fig. 3a. This ripple results from beam-energy loss during the initial period when $E_z = 0$, causing the beam in effect to be detuned. The use of a constant time lag is, of course, a worst case. A more realistic jitter model has the E_z timing error vary randomly over a scale length in z equal to λ_w . The wave phase for such a case having a root mean-square jitter of 0.1 ns is plotted in Fig. 3b and shows a phase ripple of about $\pi/8$.

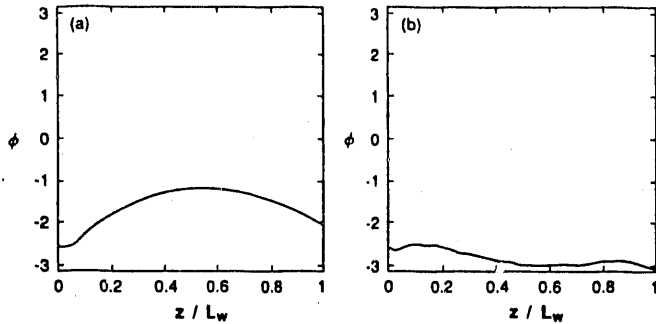


Fig. 3 Wave phase ϕ as a function of z for a beam with (a) a 0.1 ns lag in the reacceleration field and (b) a reacceleration field with an rms timing jitter of 0.1 ns.

Studies with a constant-current beam show that the final wave phase is as stable as for a beam with a linear current ramp, but there is a 10% ripple in W_{out} that persists in z . A beam with constant I_b also begins to lose particles when errors in energy or current exceed about 0.5%, indicating the reduced acceptance for this current envelope.

Conclusions

From the 1-D simulations discussed here, a standing-wave FEL appears to be a possible microwave source for a two-beam accelerator. Using a beam with modest current and energy, we find that the final microwave energy in cavities is adequate to drive a high-gradient structure, and this energy remains effectively constant in z for fluctuations in energy, current, the reacceleration field, and timing of up to 2%. The final signal phase appears to be more sensitive, with 1% errors in energy or timing causing a phase variation in z of up to $\pi/2$. Work is underway to determine how such a phase error would affect the performance of the high-gradient structure. The need for tight prebunching is another potential shortcoming of the standing-wave FEL.

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