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**Dual-Transistor Method to Determine  $\Delta V_{ot}$  and  $\Delta V_{it}$  for MOS Devices\***

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**35-Word Abstract**

It is shown how standard  $\Delta V_{th}$  and mobility measurements made on otherwise identical n- and p-channel transistors can be combined to accurately estimate radiation-induced  $\Delta V_{ot}$  and  $\Delta V_{it}$ . Applications of the method are described.

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To understand the radiation response of MOS devices, one must distinguish among effects of oxide-trapped charge and interface-trap buildup and annealing. To this end, radiation-induced threshold-voltage shifts,  $\Delta V_{th}$ , are often expressed as sums of components due to oxide-trapped charge,  $\Delta V_{ot}$ , and interface traps,  $\Delta V_{it}$  [1]. Values of  $\Delta V_{ot}$  and  $\Delta V_{it}$  commonly are estimated for MOS transistors via midgap [1,2], subthreshold slope [3,4], mobility [5-7], and/or charge-pumping [4,8] methods. Midgap, slope, and charge-pumping methods depend on accurate low-current measurements. Hence, these techniques are not well-suited for devices with large parasitic leakage. Also, because accurate low-current measurements are difficult to perform at high speed, these methods can be difficult to apply to very rapid measurements after a short radiation pulse. Mobility methods do not require low-current measurements, and so do not face these restrictions. Unfortunately, without an independent measurement of interface-trap density, mobility methods cannot be applied to one transistor to determine  $\Delta V_{ot}$  and  $\Delta V_{it}$  unambiguously [6,7].

In this summary a dual-transistor technique to estimate  $\Delta V_{ot}$  and  $\Delta V_{it}$  is described. The method is applied to n- and p-channel MOS transistors on the same chip, irradiated under identical conditions. Features of midgap and mobility methods are combined to estimate  $\Delta V_{ot}$  and  $\Delta V_{it}$  accurately and unambiguously from  $\Delta V_{th}$  and mobility measurements, made at currents 2-5 orders of magnitude above those required for typical subthreshold current or charge-pumping analysis [1-4,8], as verified in several cases.

The dual-transistor method requires that otherwise identical n- and p-channel transistors (e.g. on the same wafer, to best match oxide radiation response) be irradiated under the same conditions at the same oxide electric field. With this pre-condition, it is assumed: (i) that  $\Delta V_{ot}$  is approximately equal for these n- and p-channel transistors; (ii) that interface traps predominantly are charged negatively for n-channel and positively for p-channel devices; (iii) that the postirradiation mobility,  $\mu$ , is related (approximately) to the preirradiation mobility,  $\mu_0$ , and the interface-trap buildup,  $\Delta N_{it}$ , by the Sun-Plummer equation [5],  $\mu/\mu_0 \approx [1 + \alpha(\Delta N_{it})]^{-1}$ , where (iv)  $\alpha$  is taken to be the same for n- and p-channel transistors. Assumption (i) is consistent with MOS-capacitor work where  $\Delta V_{ot}$  was found not to depend on doping type or level [1]; (ii) is similar to, but less restrictive than, the common assumption of interface-trap charge neutrality at midgap [1,2]; and (iii) and (iv) are consistent with extensive data of Sexton and Schwank [7].

With these plausible assumptions, equations can be derived that express  $\Delta V_{ot}$  and  $\Delta V_{it}$  in terms of n- and p-channel threshold-voltage shifts,  $\Delta V_{thn}$  and  $\Delta V_{thp}$ , and mobility degradation,  $\mu_n/\mu_{n0}$  and  $\mu_p/\mu_{p0}$ . To do so,  $\Delta V_{thn}$  and  $\Delta V_{thp}$  are first parameterized in terms of variables representing  $\Delta V_{ot}$  and  $\Delta V_{it}$ :

$$(1) \Delta V_{thn} = -H + S_n \quad \text{and} \quad (2) \Delta V_{thp} = -H - S_p.$$

Here, by (i),  $\Delta V_{otn} \approx \Delta V_{otp} = -H$ . By (ii),  $\Delta V_{itn} = S_n$  and  $\Delta V_{itp} = -S_p$ , where the new variables  $H$ ,  $S_n$ , and  $S_p$  are positive by definition. Because different parts of the bandgap are being sampled [1,4,8],  $S_n$  is generally not equal to  $S_p$ . From (iii) and (iv), with a further variable change, one may re-write the Sun-Plummer equations to complete a system of 4 equations in 4 unknowns:

$$(3) S_n = \beta_n/\alpha^* \quad \text{and} \quad (4) S_p = \beta_p/\alpha^*.$$

Here  $\alpha^* S_n = \alpha \Delta N_{itn}$ , and  $\alpha^* S_p = \alpha \Delta N_{itp}$ , where  $\alpha$  is the usual Sun-Plummer constant [5,7]. Also,  $\beta_n = (\mu_{on}/\mu_n)^{-1}$ , and  $\beta_p = (\mu_{op}/\mu_p)^{-1}$ , where  $\mu_{on}$  and  $\mu_{op}$  are n- and p-channel preirradiation mobilities, and  $\mu_n$  and  $\mu_p$  are postirradiation mobilities, respectively.

Equations (1)-(4) can be solved exactly to arrive at the final expressions, in terms of the quantities of interest:

$$\Delta V_{itn} = \frac{\beta_n (\Delta V_{thn} - \Delta V_{thp})}{\beta_n + \beta_p}, \quad (5)$$

$$\Delta V_{itp} = -\frac{\beta_p (\Delta V_{itn})}{\beta_n} = -\frac{\beta_p (\Delta V_{thn} - \Delta V_{thp})}{\beta_n + \beta_p}, \quad (6)$$

$$\Delta V_{otn} = \Delta V_{otp} = \frac{\beta_p (\Delta V_{thn}) + \beta_n (\Delta V_{thp})}{\beta_n + \beta_p}, \quad (7)$$

and

$$\alpha^* = \frac{\beta_p + \beta_n}{\Delta V_{thn} - \Delta V_{thp}}. \quad (8)$$

Note that the value of the constant  $\alpha^*$ , that relates  $\mu$  to  $\Delta V_{it}$ , emerges naturally in the solution. Since  $\alpha$  should be approximately constant (e.g. through a series of irradiations) if Eqs. (3) and (4) are valid, Eq. (8) can be used as a "self-consistency" check of the method.

The dual-transistor method is compared to midgap and subthreshold slope methods below. For the dual-transistor analysis, n- and p-channel transistor threshold voltages and mobilities were measured at  $\pm 5$  V drain voltage and  $\pm 16$ -100  $\mu$ A channel currents [9]. Subthreshold slope and midgap methods were applied as discussed in Refs. 1 and 3.

To see how the methods compare for cases in which midgap and subthreshold slope techniques are expected to be valid [10], consider the following example. In Fig. 1(a) current-voltage (I-V) characteristics are shown for n-channel transistors with 47-nm oxides as a function of x-ray dose. Values of  $\Delta V_{itn}$  inferred from these curves with the dual-transistor, midgap, and slope methods are shown in Fig. 1(b). To within the uncertainties of the midgap and slope methods, values of  $\Delta V_{itn}$  inferred with all three methods agree to within better than  $\pm 10$  percent. (Uncertainties in midgap and slope analyses are caused by the slope change that occurs at  $I_{DS} \approx 10$  nA at the highest doses in Fig. 1(a), reflected by the error bars in Fig. 1(b)). Similar agreement is observed for  $\Delta V_{otn}$ ,  $\Delta V_{otp}$ , and  $\Delta V_{itp}$ . Further, for these data, the value of  $\alpha$  inferred from dual-transistor analysis [Eq. (8)] is  $0.50 \pm 0.03$ , demonstrating self-consistency through the irradiation series. Similar agreement among the methods is also found for irradiation and anneal of two other types of devices, up to 10-times more radiation resistant [11].

Now that the dual-transistor method has been verified in these cases, consider the following example in which the application of midgap and slope methods is questionable because of high leakage currents. In Figs. 2(a) I-V curves and (b) inferred values of  $\Delta V_{itn}$  are shown as a function of Co-60 dose and elevated-temperature anneal for n-channel transistors with large parasitic leakage. Previous applications of midgap and slope analyses [1,3] have suggested that the 1-2 decades of "straight" I-V curve above the leakage in Fig. 2(a) would allow for reasonable estimates of  $\Delta V_{ot}$  and  $\Delta V_{it}$ . Along these lines, midgap and slope values shown in Fig. 2(b) are derived [1,3] from current measurements of 20-100 nA to try to minimize the impact of the leakage. Still, for doses  $\geq 50$  krad, values of  $\Delta V_{itn}$  obtained with these

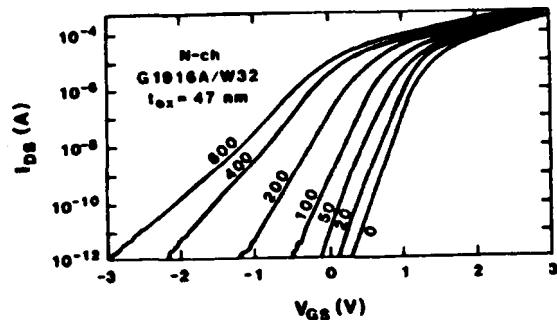
methods are up to  $\sim 4$  times higher than values obtained with the dual-transistor method. After 1-h anneal at 100°C, the parasitic leakage [Fig. 2(a)] is reduced to  $\leq 10^{-11}$  A. Thus, midgap and subthreshold slope methods are expected to be valid through anneal. And, in Fig. 2(b), excellent agreement among all methods is observed through this region. The value of  $\alpha^*$  [Eq. (8)] is  $1.25 \pm 0.11$  through irradiation and anneal, once again showing the self-consistency of the dual-transistor analysis. Also, values of  $\Delta V$  obtained via dual-transistor analysis increase linearly with dose [Fig. 2(b)], as expected [12], in contrast to the strong nonlinearity suggested by midgap and slope analyses. Finally, for the corresponding p-channel transistors that did not show high leakage, values of  $\Delta V_{otp}$  and  $\Delta V_{itp}$  agree to within better than  $\pm 10\%$  for all methods. It is concluded that the dual-transistor method can provide much more accurate estimates of  $\Delta V_{ot}$  and  $\Delta V_{it}$  than subthreshold current (and charge-pumping) methods for devices with high leakage.

Like single-transistor techniques, the dual-transistor method has limitations. It cannot be applied if otherwise identical n- and p-channel transistors are not available (preferably on the same wafer). No information is given about changes in interface-trap density through the bandgap. Also, the method is difficult to apply at small electric fields, where work function differences between n- and p-channel transistors may not be known accurately enough to permit compensation for the resulting differences in radiation response. However, the dual-transistor method does provide accurate estimates of  $\Delta V_{ot}$  and  $\Delta V_{it}$  in cases of great interest, even when large parasitic leakage (e.g. due to parasitic field leakage or high-temperature operation) is present. It is also much easier to apply than subthreshold current or charge-pumping techniques at very short times after a radiation pulse, as shown in detail in the full paper. The dual-transistor method therefore fills important gaps in the characterization of MOS radiation response.

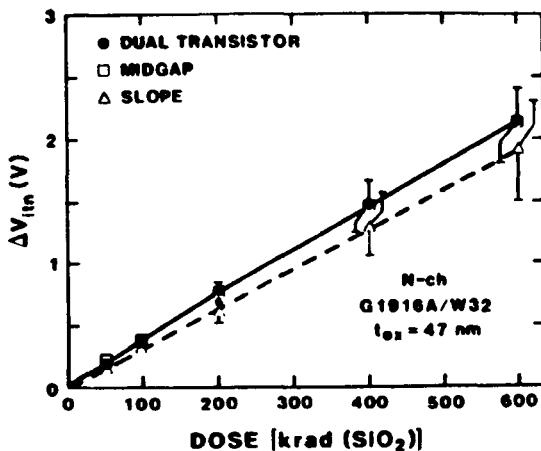
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9. In mobility methods, the "dc-conductance" mobility (Refs. 5-7) is usually measured, instead of the "field-effect" mobility used here. Dual-transistor analysis may also be applied to linear threshold-voltage and/or dc-conductance mobility measurements. Saturation values are shown here to illustrate the (perhaps surprising) degree of accuracy that can be obtained even with simpler high-current measurements.
10. From previous work (e.g. Ref. 8), charge-pumping is expected to provide a similar estimate of  $\Delta V_{it}$  in this case.
11. D. M. Fleetwood and M. R. Shaneyfelt, unpublished.
12. D. M. Fleetwood, P. S. Winokur, and J. R. Schwank, IEEE Trans. Nuc. Sci. NS-35, 1497 (1988).

(a)

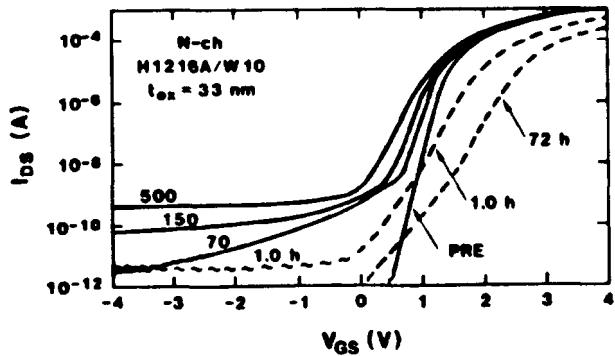


(b)

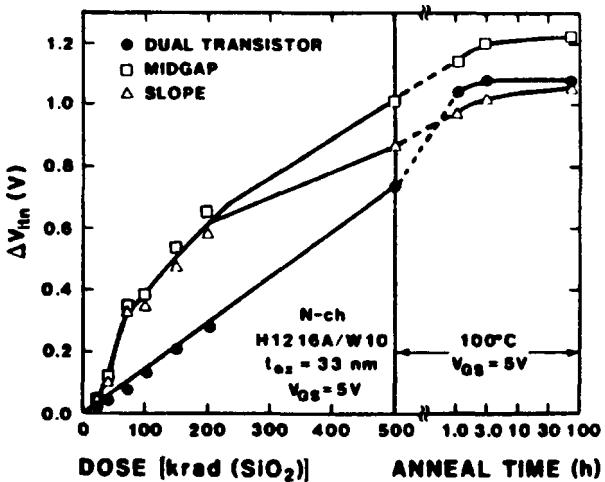


1. (a) Current-voltage traces and (b) inferred values of  $\Delta V_{itn}$  for  $3 \times 16 \mu\text{m}$  transistors with 47-nm oxides. The total x-ray dose received before each trace in (a) is shown in krad( $\text{SiO}_2$ ). The radiation bias was  $V_{GS} = +6 \text{ V}$ .

(a)



(b)



2. (a) Representative current-voltage traces and (b) inferred values of  $\Delta V_{itn}$  for  $2 \times 16 \mu\text{m}$  transistors with 33-nm oxides. The Co-60 dose or the annealing time at  $100^\circ\text{C}$  is shown in (a) for each trace. The irradiation and anneal bias was  $V_{GS} = +5 \text{ V}$ .