

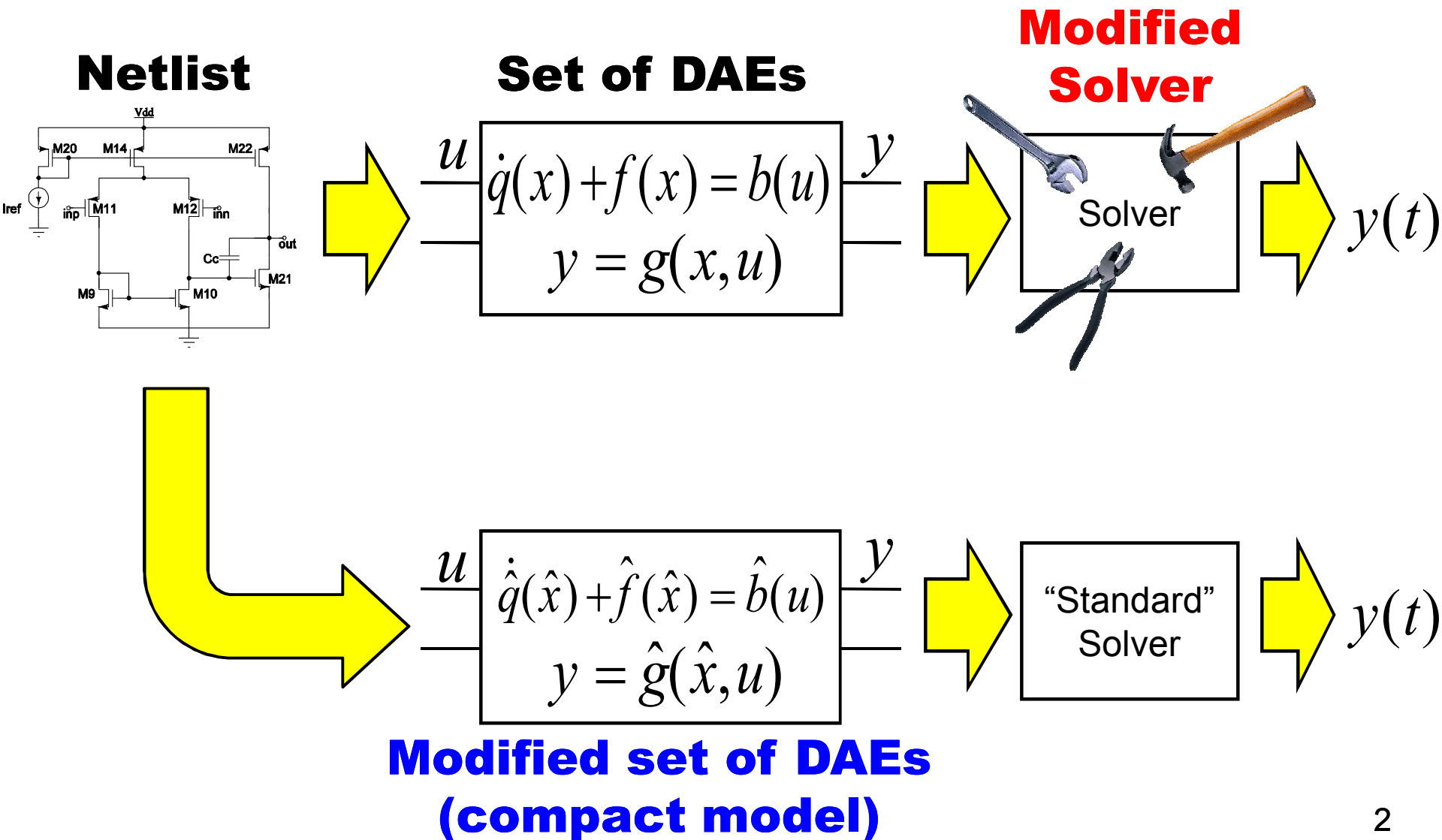
Accelerated Simulation of Analog Systems Using Linear and Nonlinear Robust Compact Models

Brad Bond

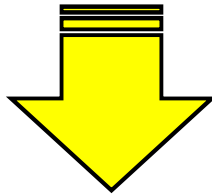
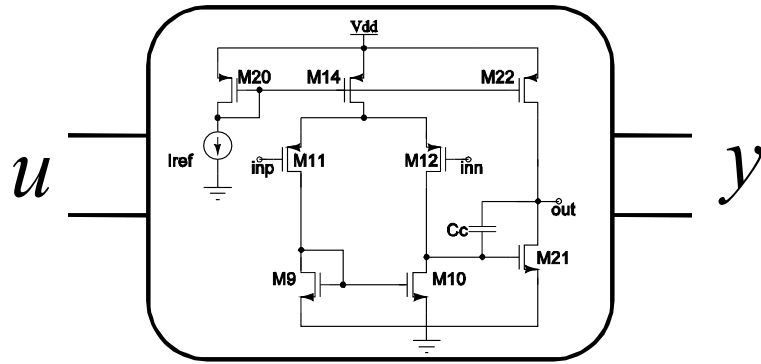
Sandia National Labs, Albuquerque NM
February 28, 2011

Thanks to: Luca Daniel, Alex Megretski, Zohaib Mahmood,
Yan Li, Tarek El-Moselhy -- MIT, Cambridge MA

Accelerated Simulation of Analog Systems Using Linear and Nonlinear Robust Compact Models



Accelerated Simulation of Analog Systems Using Linear and Nonlinear Robust **Compact Models**



$$\begin{aligned} \dot{\hat{q}}(\hat{x}) + \hat{f}(\hat{x}) &= \hat{b}(u) \\ y &= \hat{g}(\hat{x}, u) \end{aligned}$$

Compact Model

- Port description of system
 - Device, block, circuit
- Fits into solver framework
 - System of ODEs or DAEs
- Reduced complexity
 - Faster function evaluations and system solves

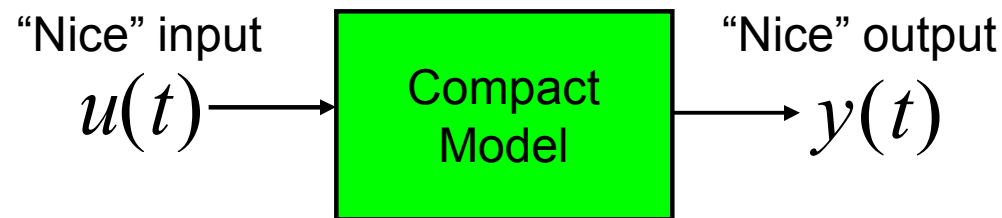
~~“Model Order Reduction”~~
~~“Reduced Order Model”~~

Accelerated Simulation of Analog Systems Using Linear and Nonlinear **Robust** Compact Models

Accurate

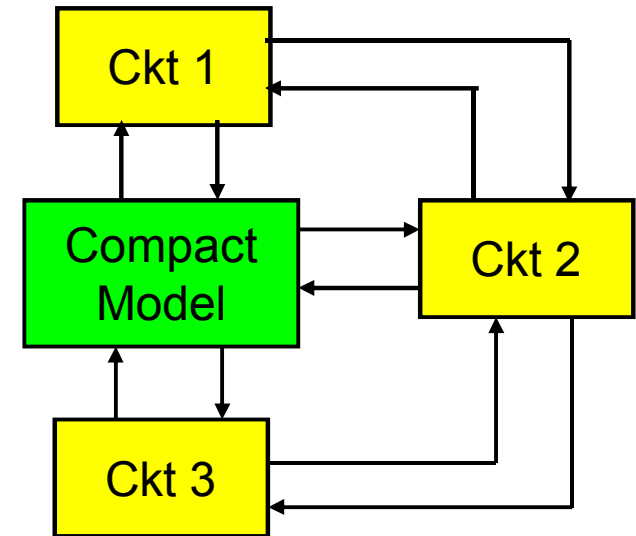
- Guarantee of accuracy between original model and compact model
- Wide range of inputs (amplitude, frequency, shape, ...)

Stable

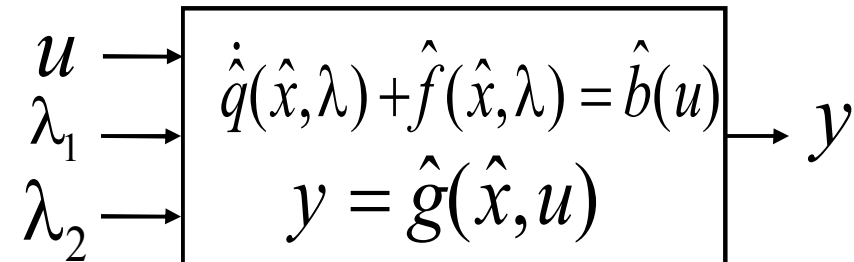


- Bounded input → bounded output
- Small input perturbation produces small output perturbation
- **Difficult to enforce global stability via projection for nonlinear system**

Passive

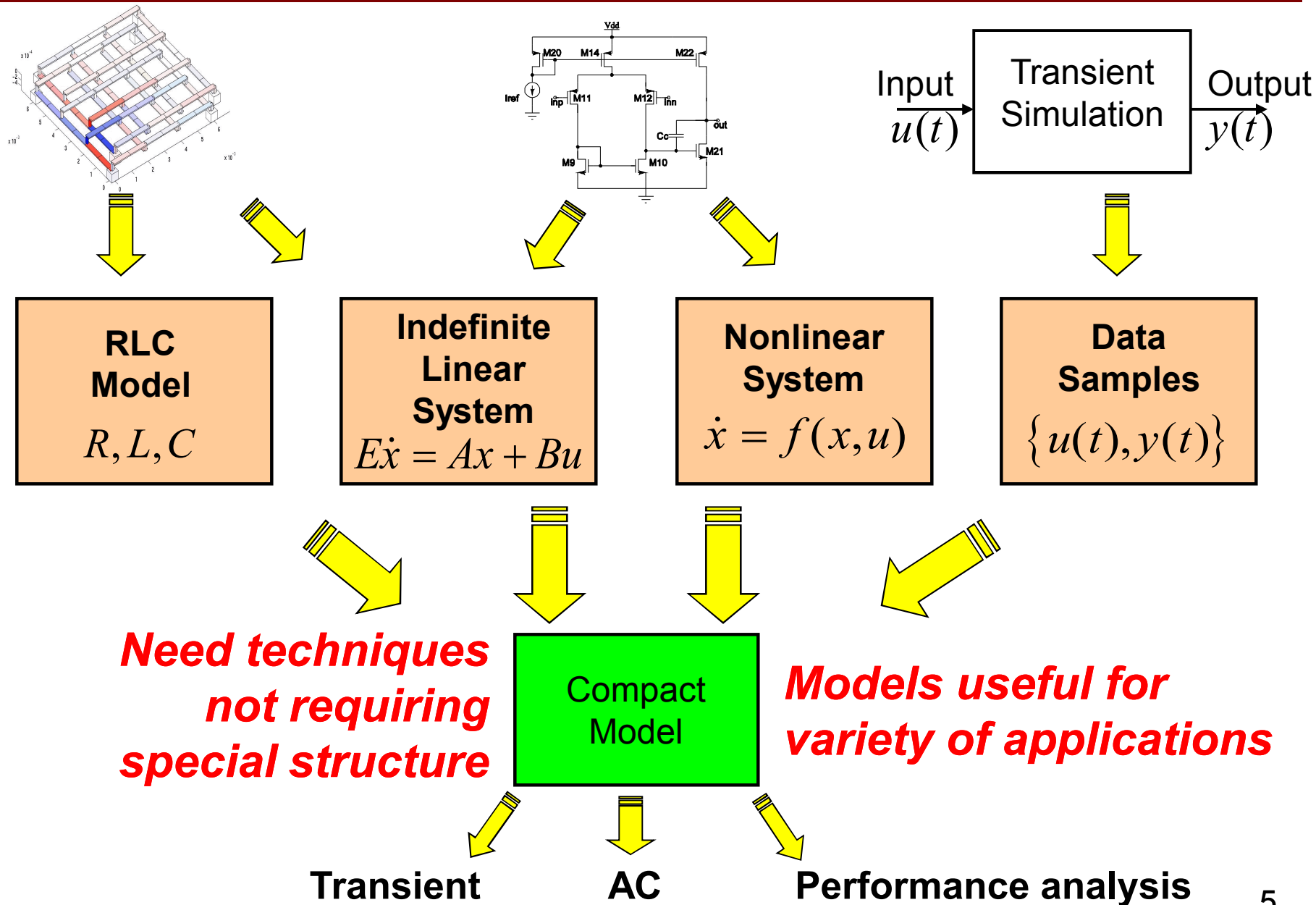


Parameterized



(e.g. design or manufacturing parameters)

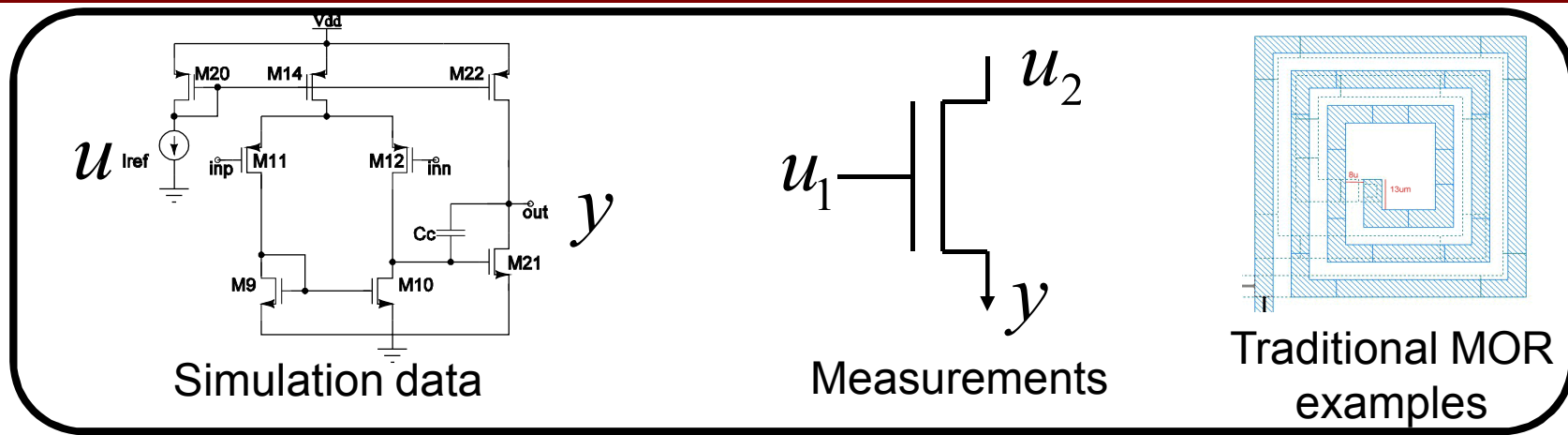
Accelerated Simulation of *Analog Systems* Using Linear and Nonlinear Robust Compact Models



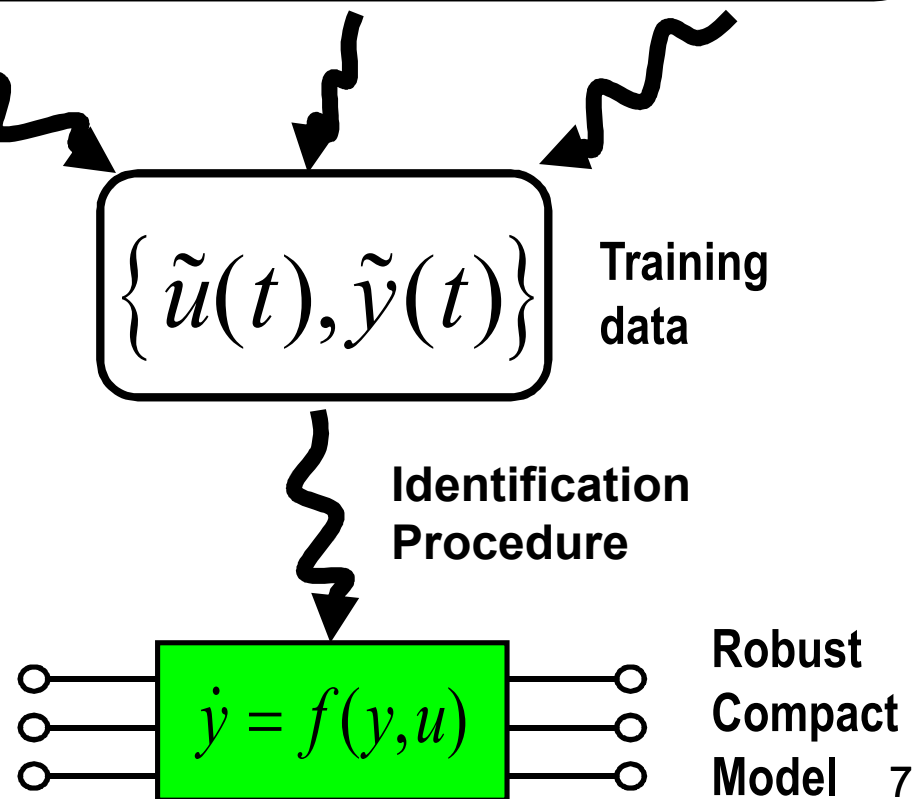
Outline

- Motivation
- System Identification (SYSID): Introduction
- SYSID: Theory
- SYSID: How to make it practical
- Examples

SYSID Intro (1): System Identification



- Solves MOR problems
 - Does not require system equations
 - Can enforce stability by construction



SYSID Intro (2): Model Terminology / Notation

		Continuous time	Discrete time
First order	Explicit I/O system	$\dot{y} = f(y, u)$	$y_t = f(y_{t-1}, u_{t-1})$
	Implicit I/O system	$F(\dot{y}, y, u) = 0$	$F(y_t, y_{t-1}, u_t) = 0$
Higher order	Delays, derivatives	$F(y, \dot{y}, \ddot{y}, \dots, u, \dot{u}, \ddot{u}, \dots) = 0$	$F(y_t, y_{t-1}, \dots, y_{t-m}, u_t, u_{t-1}, \dots, u_{t-k}) = 0$
	States	$F(\dot{x}, x, u) = 0$ $G(y, x) = 0$	$F(x_t, x_{t-1}, u_t) = 0$ $G(y_t, x_t) = 0$

f -- Explicit system

u -- Inputs

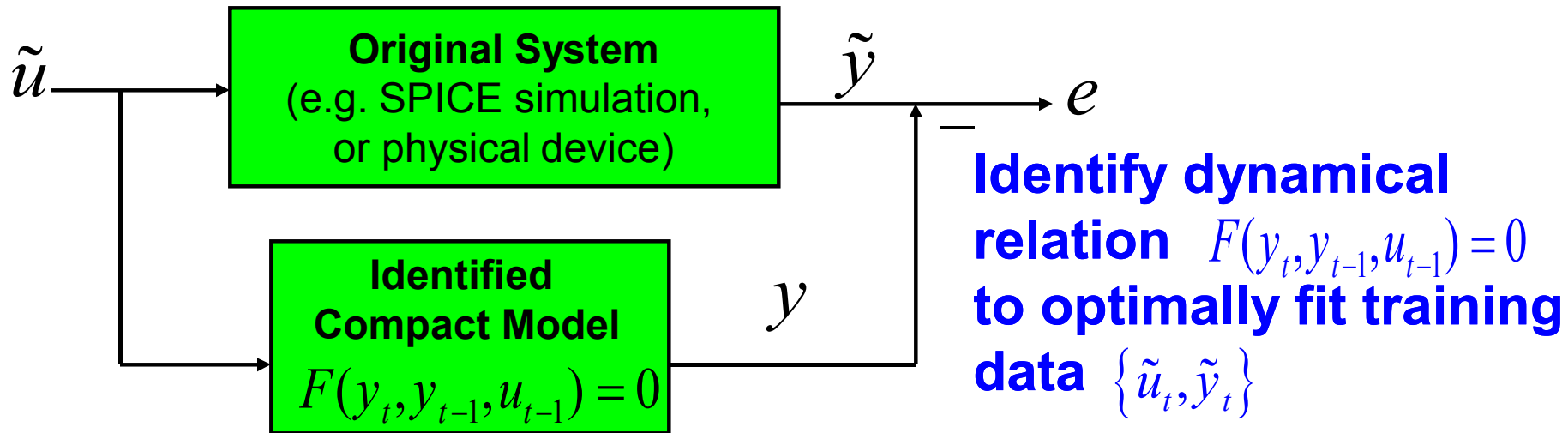
y -- Outputs

F -- Implicit system

(e.g. $F = \dot{y} - f(y, u) = 0$)

x -- States (optional)

SYSID Intro (3): Problem Statement



System Identification.... Is it just Fitting?

$$\min_F \sum_t |\mathbb{E}(F, \tilde{y}, \tilde{u})|^2$$

s.t. Robustness(F)

- What to minimize?
- Robustness constraint?
- How to describe F ?

Optimization problem must be nice

Outline

- Motivation
- System Identification (SYSID): Introduction
- SYSID: Theory
- SYSID: How to make it practical
- Examples

SYSID (1): Nonlinear Model Description

Implicit DT Models

$$F(y_t, y_{t-1}, u_t) = 0$$

Unlike projection, no equations to start from

Basis Functions

$$F(y_t, y_{t-1}, u_t) = \sum_j \alpha_j \phi_j(y_t, y_{t-1}, u_t)$$

Polynomial basis

$$\phi_j = y_t^{a_j} y_{t-1}^{b_j} u_t^{c_j}$$

Implicit model captures highly nonlinear behavior

For example,

$$y_t^2 - y_{t-1} + u_t^3 = 0$$



$$y_t = \sqrt{y_{t-1} - u_t^3}$$

$$q(y_{t-1})y_t - p(y_{t-1}, u_t) = 0$$



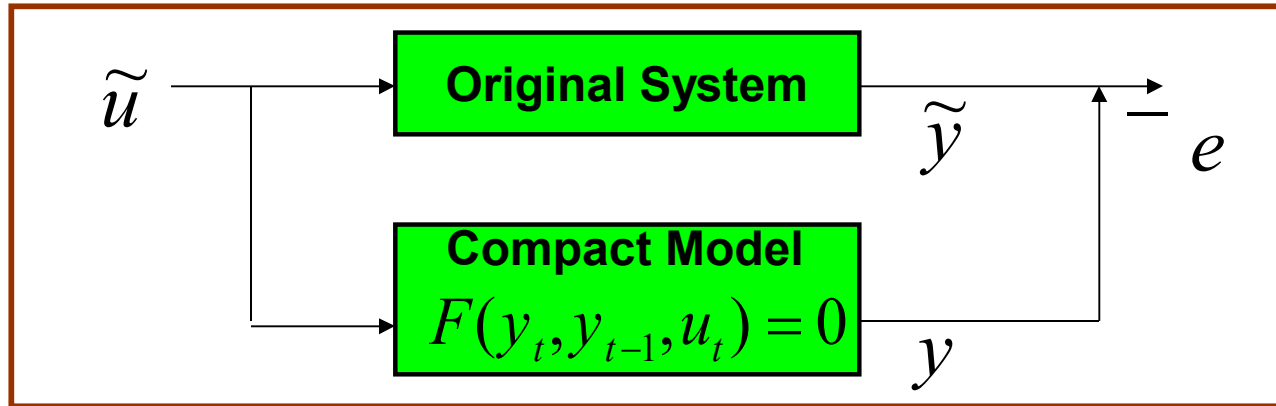
$$y_t = \frac{p(y_{t-1}, u_t)}{q(y_{t-1})}$$

**Rational
Models**

Later in the talk

Models with states, higher order systems, continuous time ¹¹

SYSID (2): Explicit Error Minimization



Goal:
minimize $\sum_t |e_t|^2$

Assume explicit system
 $y_t = f(y_{t-1}, u_{t-1})$

$$|e_t| = |\tilde{y}_t - y_t| = |\tilde{y}_t - f(y_{t-1}, u_t)|$$

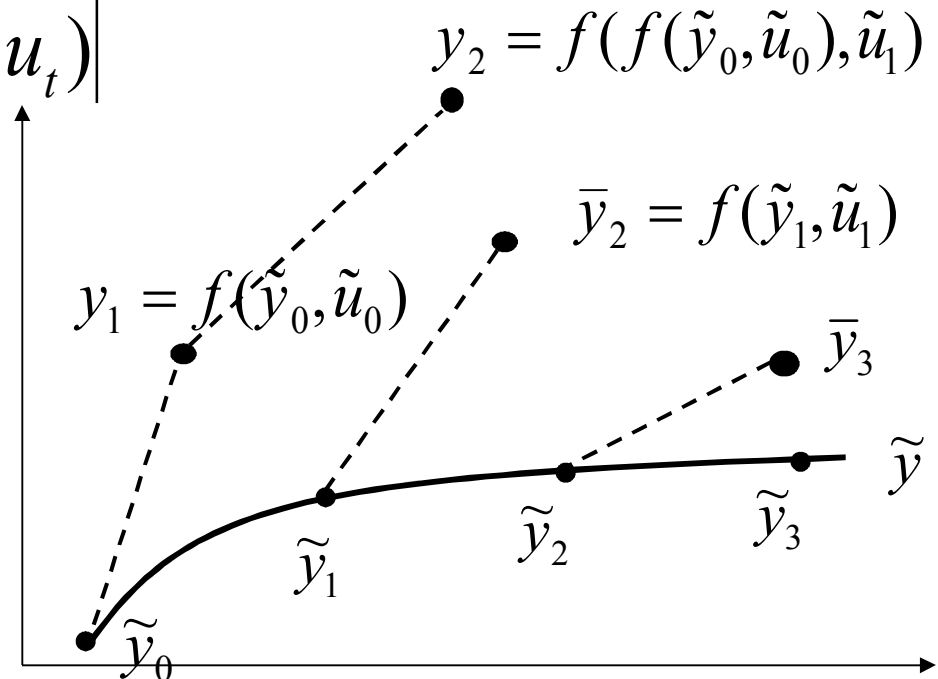
$$|e_1| = |\tilde{y}_1 - f(\tilde{y}_0, \tilde{u}_0)|$$

~~$$|e_2| = |\tilde{y}_2 - f(f(\tilde{y}_0, \tilde{u}_0), \tilde{u}_1)|$$~~

Nonlinear in unknown coefficients

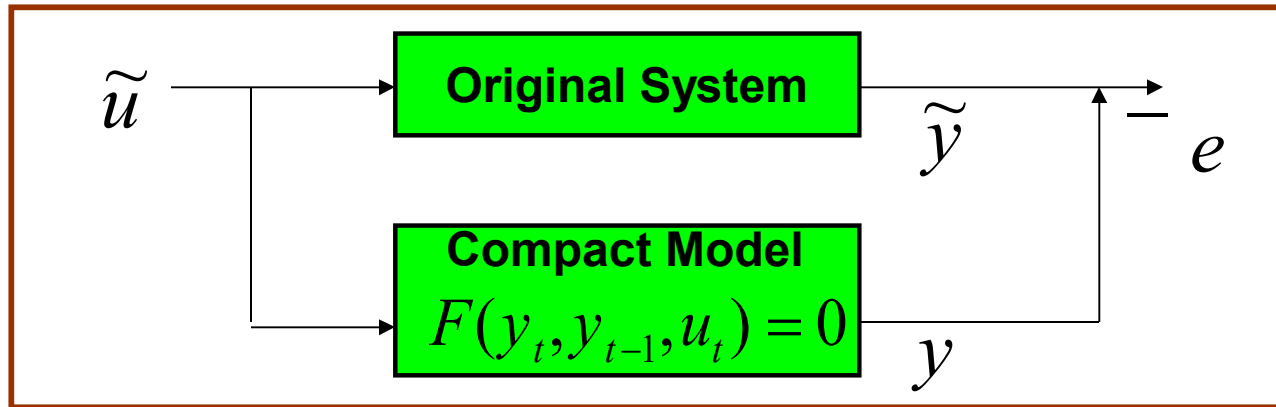
$$|e_2| = |\tilde{y}_2 - f(\tilde{y}_1, \tilde{u}_1)|$$

Can only minimize one-step errors



Must guarantee errors do not accumulate

SYSID (3): Implicit Error Minimization



Goal:

$$\text{minimize } \sum_t |e_t|^2$$

- For implicit system, minimize equation error (i.e. residual)

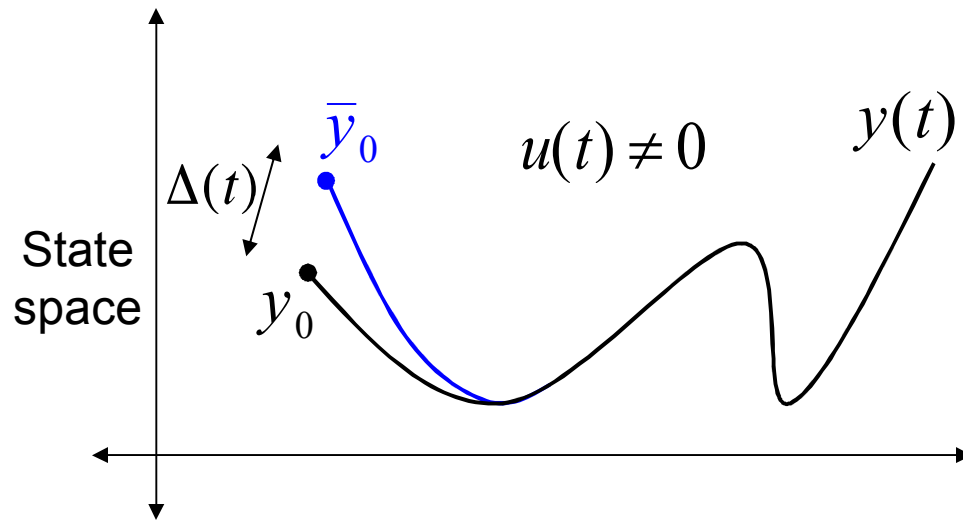
$$F(y_t, y_{t-1}, u_{t-1}) = 0 \quad \Rightarrow \quad \text{minimize } \sum_t |F(\tilde{y}_t, \tilde{y}_{t-1}, \tilde{u}_t)|^2$$

- Small equation error doesn't always imply small output error

$$\begin{array}{l} Ax_1 = b_1 \\ Ax_2 = b_2 \end{array} \quad \|b_1 - b_2\| \text{ small} \not\Rightarrow \|x_1 - x_2\| \text{ small}$$

Need relation between equation error and output error

SYSID (4): Incremental Stability



One possible constraint –
sufficient, but not necessary

- Perturbations to solutions decay
- Errors do not accumulate

Incremental stability constraint

$$(y - \bar{y})^T (F(y, u) - F(\bar{y}, u)) \geq h_+(y, \bar{y}) - h_-(y, \bar{y}) + |y - \bar{y}|^2$$

$h(y, \bar{y})$ -- Certificate of stability

- Equation error bounds output error

$$\sum_t |F(\tilde{y}_t, \tilde{y}_{t-1}, \tilde{u})|^2 \geq \sum_t |y_t - \tilde{y}_t|^2$$

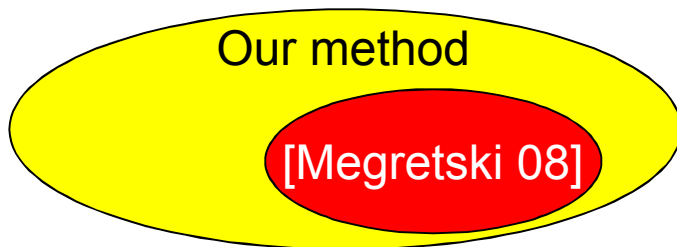
Modified incremental stability constraint

[Bond et al TCAD10]

$$-2\Delta_+^T \dot{F}(y, u) \Delta + h(y_+, \Delta_+) - h(y_-, \Delta_-) \leq -|\Delta_+|^2$$

- Semidefinite constraint
- Tighter error bound
- Larger class of admissible stable models F

Space of Stable F



SYSID (5): Identification Procedure

Optimization Problem

$$\min_{F,r,h} \sum_t r(\tilde{y}_t, \tilde{u}_t) \quad \text{subject to}$$

Robust equation error, upper bounds linearized output error

$$r_t + 2\Delta_0^T (F_t + \dot{F}_t \Delta) - |\Delta|^2 + h_- - h_+ \geq 0, \quad \forall t, \Delta$$

Robust dissipation constraint, implies local incremental stability

$$2\Delta_0^T \dot{F}(y, u) \Delta - |\Delta|^2 + h_- - h_+ \geq 0, \quad \forall u, y, \Delta$$

Optional constraint, enforces global incremental stability

Semidefinite constraints

Inputs

$\{\tilde{u}_t, \tilde{y}_t\}$ - Training data set

$\Phi = \{\phi_j(y, u)\}$ - Set of basis functions describing F, r, h

Outputs

- Coefficients defining**

F - Stable reduced model

h - Storage function, certifies stability

r - Error bound

$$F(y, u) = \sum_j \alpha_j \phi_j(y, u)$$

Outline

- Motivation
- System Identification (SYSID): Introduction
- SYSID: Theory
- SYSID: How to make it practical
- Examples

SYSID (1): Complexity

- **Most general model**

$$F(\lambda, x_t, \dots, x_{t-m}, u_t, \dots, u_{t-k}) = 0 \qquad G(x_t, y_t) = 0$$

- **Complexity (# of coefficients) determined by**
 - Number of inputs, outputs, states, parameters
 - Number of delays (input and state memory)
 - Polynomial degrees

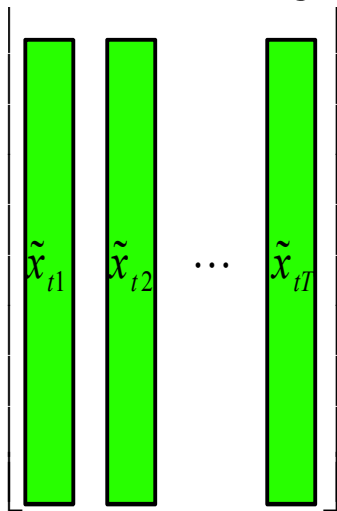
$$\phi = \phi(\lambda, x_t, \dots, x_{t-m}, u_t, \dots, u_{t-k}, y_t, \dots, y_{t-l}) = \prod_{h,i,j,k,l,m,n,r,s} u_{t-i,n}^{p_j} x_{t-j,m}^{p_k} y_{t,s}^{p_h} \lambda_r^{p_l}$$

- **Need to keep total number of coefficients small**

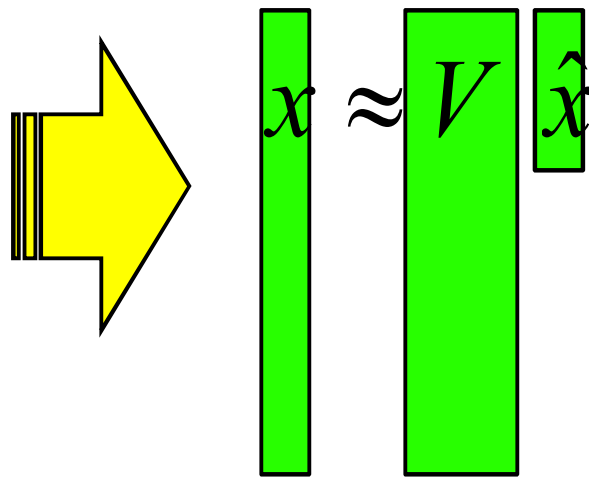
SYSID (2): Projecting States

- Model with large number of states: $F(x_t, x_{t-1}, u_t) = 0$

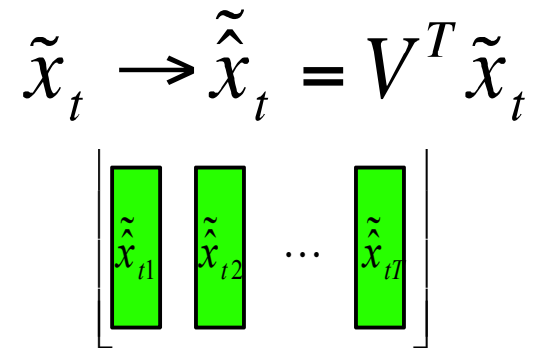
Generate training data



Find low-order basis



Project state samples



- Fit to reduced set of states: $F(\hat{x}_t, \hat{x}_{t-1}, u_t) = 0$

- Connections to traditional projection approaches**

$$U^T \tilde{F}(V\hat{x}, u) = 0 \quad \xrightarrow{\quad} \quad F(\hat{x}, u) = 0 \quad \xleftarrow{\quad} \quad \min_F E(F(\hat{x}, u))$$

- Equations obtained by fitting to data rather than from residual orthogonality condition

SYSID (3): Reduced Basis

- Model with large number of delays

$$F(\hat{x}_t, \dots, \hat{x}_{t-m}, u_t, \dots, u_{t-k}) = 0$$

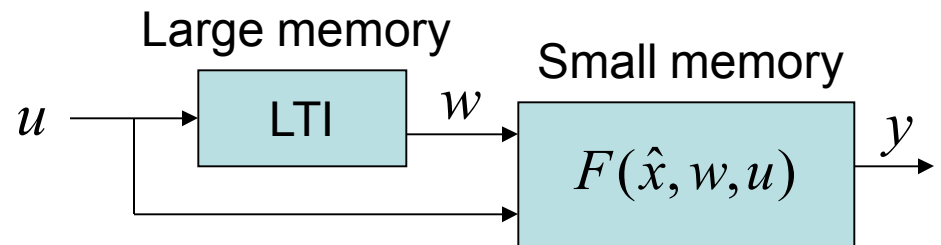
- Many basis terms for large polynomial degree

$$\left(\hat{x}_{t-1}, \dots, \hat{x}_{t-m}, u_t, \dots, u_{t-k}\right)^p$$

- Fit to reduced set of basis terms
 - Linear transformation of basis

$$u_t^2 + u_{t-1}^2 + u_{t-2}^2 + 2u_t u_{t-1} + 2u_t u_{t-2} + 2u_{t-1} u_{t-2} = \left(\overbrace{u_t + u_{t-1} + u_{t-2}}^{w_t}\right)^2$$

$$w_t = \sum_j \beta_j u_{t-j}$$



SYSID (4): Algorithm

Model complexity determined by m, k, p, q

$$y_t = \frac{p(y_{t-1}, \dots, y_{t-m}, u_t, \dots, u_{t-k})}{q(y_{t-1}, \dots, y_{t-m}, u_t, \dots, u_{t-k})}$$

Select model complexity: $\{m, k, p, q\}$

Space of Rational models
with m', k', p', q'

Space of Rational models
with m, k, p, q

Locally stable rational
models with m, k, p, q

Globally stable rational
models with m, k, p, q

Increase model
complexity: $\{m, k, p, q\}$

Fit without stability
constraint

Poor fit

Good fit

Increase
 $\{m, k, p, q\}$

Enforce local stability

Poor fit

Good fit

Increase
 $\{m, k, p, q\}$

Enforce global stability

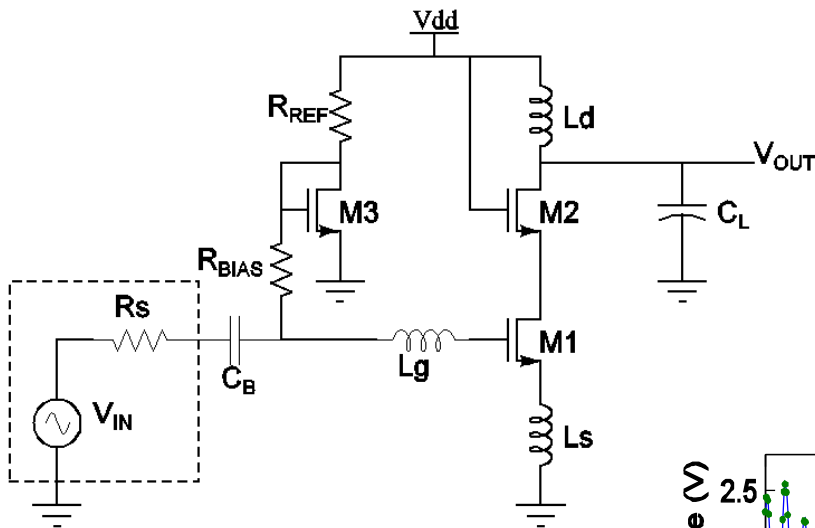
Poor fit

Good fit

Outline

- Motivation
- System Identification (SYSID): Introduction
- SYSID: Theory
- SYSID: How to make it practical
- Examples

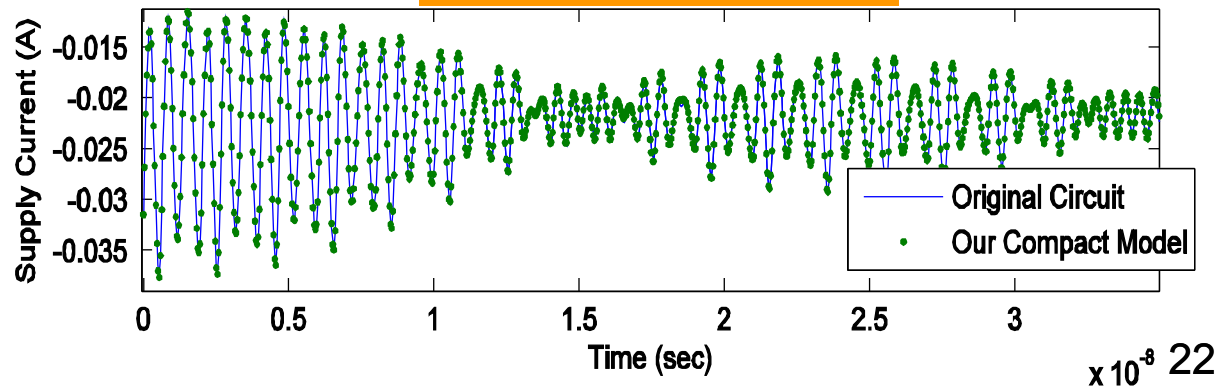
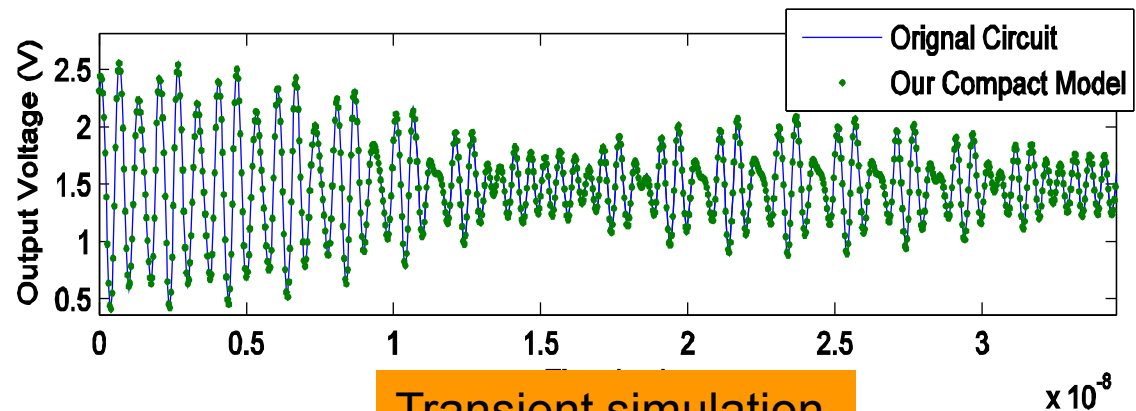
Example: Low Noise Amplifier (1)



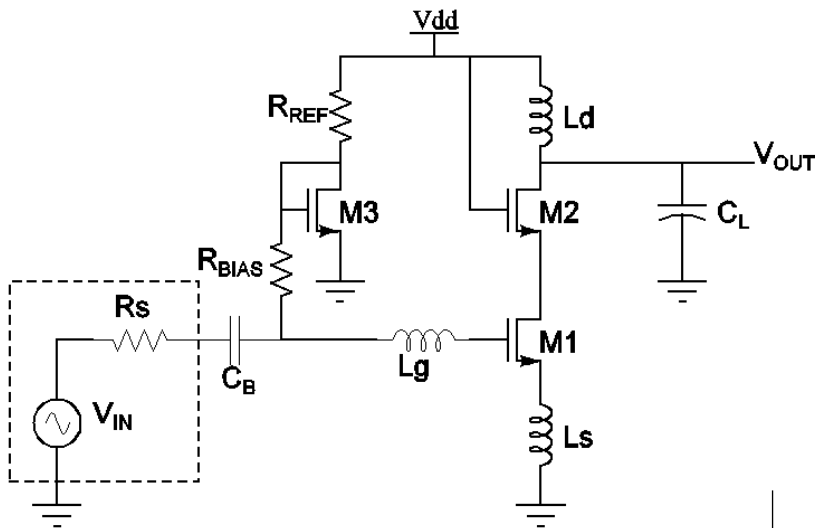
$$Q_2(y_{t-1}, u_t, u_{t-1})y_t = p_3(y_{t-1}, u_t, u_{t-1})$$

$$Q = \begin{bmatrix} Q_{11} & Q_{12} \\ Q_{21} & Q_{22} \end{bmatrix} \quad p = \begin{bmatrix} p_1 \\ p_2 \end{bmatrix}$$

- Low noise amplifier
- Identified rational model with feedback
- Model has two outputs
- 102 coefficients describing model



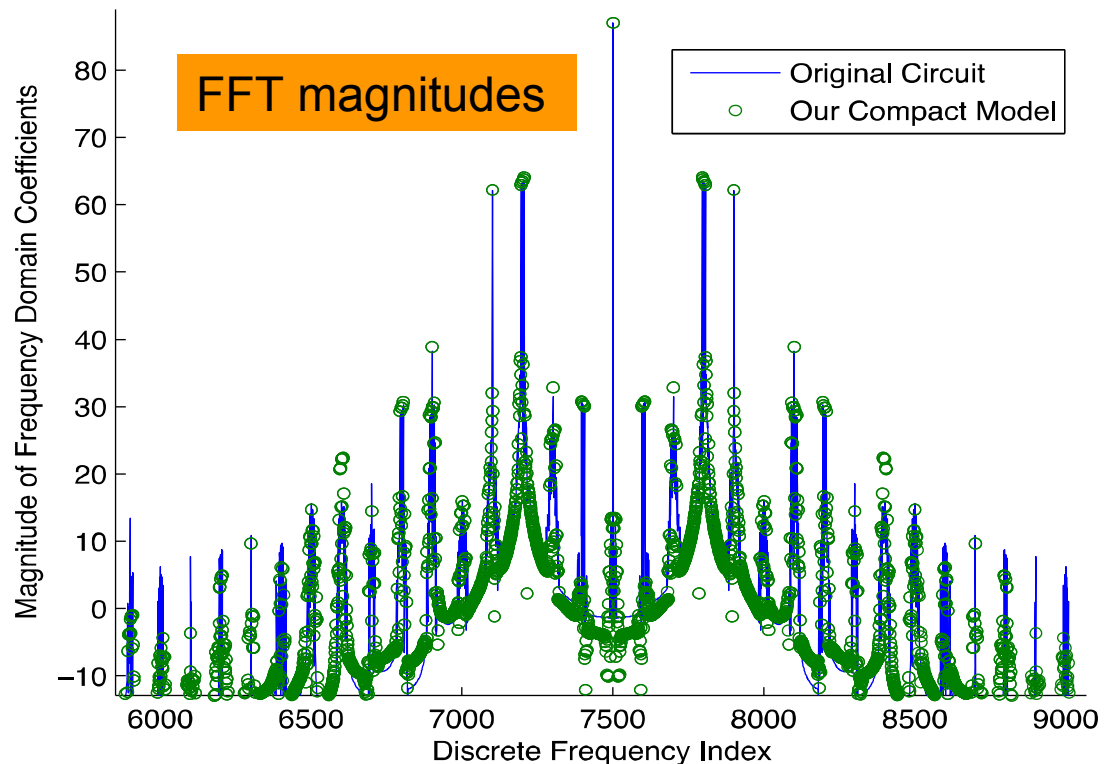
Example: Low Noise Amplifier (2)



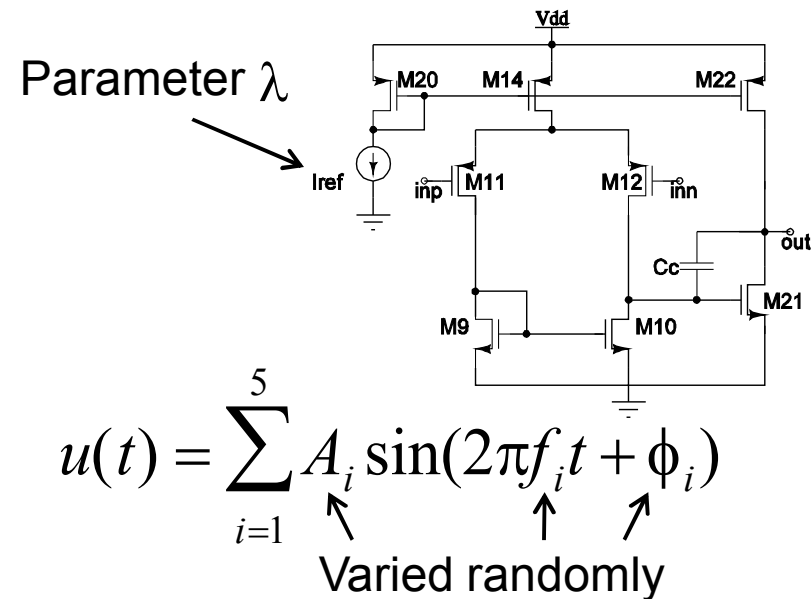
$$Q_2(y_{t-1}, u_t, u_{t-1})y_t = p_3(y_{t-1}, u_t, u_{t-1})$$

$$Q = \begin{bmatrix} Q_{11} & Q_{12} \\ Q_{21} & Q_{22} \end{bmatrix} \quad p = \begin{bmatrix} p_1 \\ p_2 \end{bmatrix}$$

- Low noise amplifier
- Identified rational model with feedback
- Model has two outputs
- 102 coefficients describing model



Example: Parameterized System – Opamp



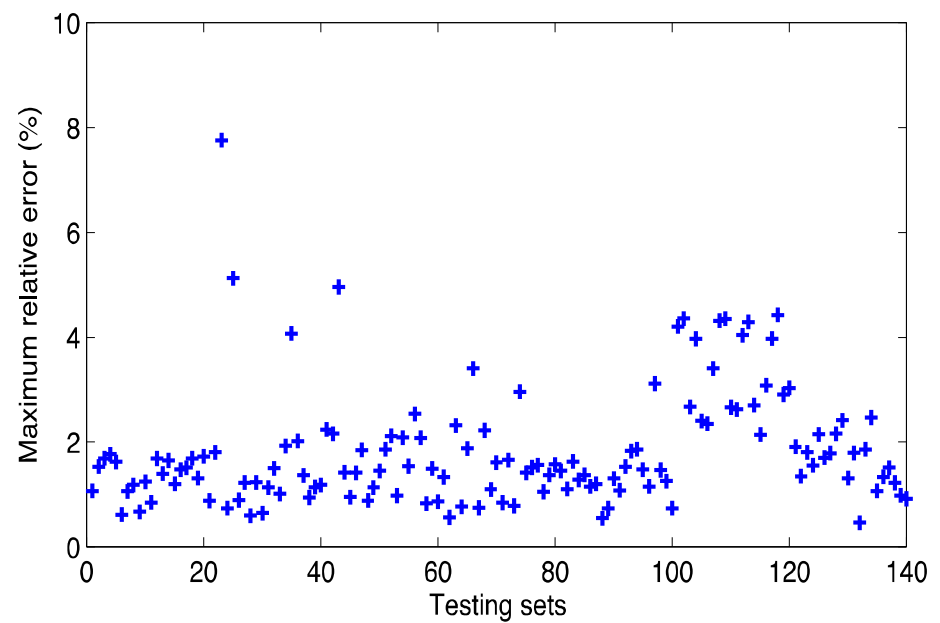
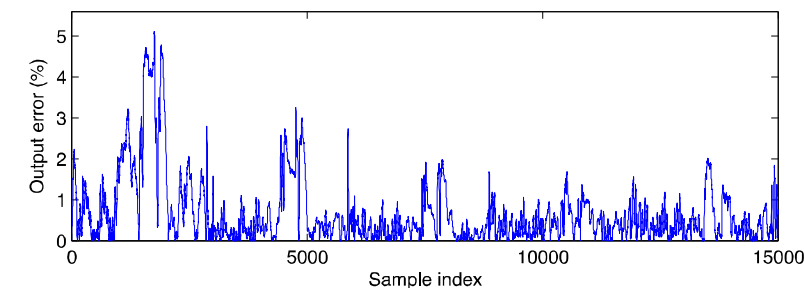
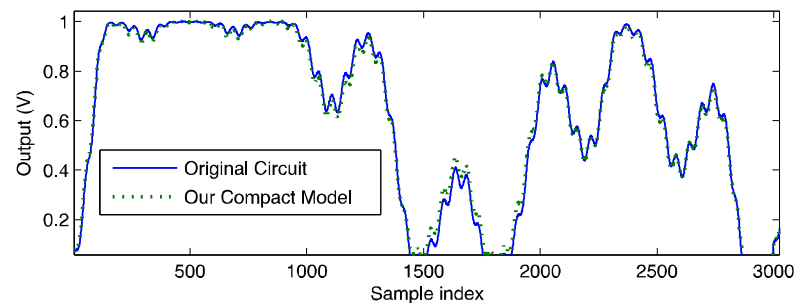
$$y_t = \frac{p_3(\lambda, y_{t-1}, u_t, u_{t-1})}{q_4(y_{t-1}, u_t, u_{t-1})}$$

97 coefficients describing model

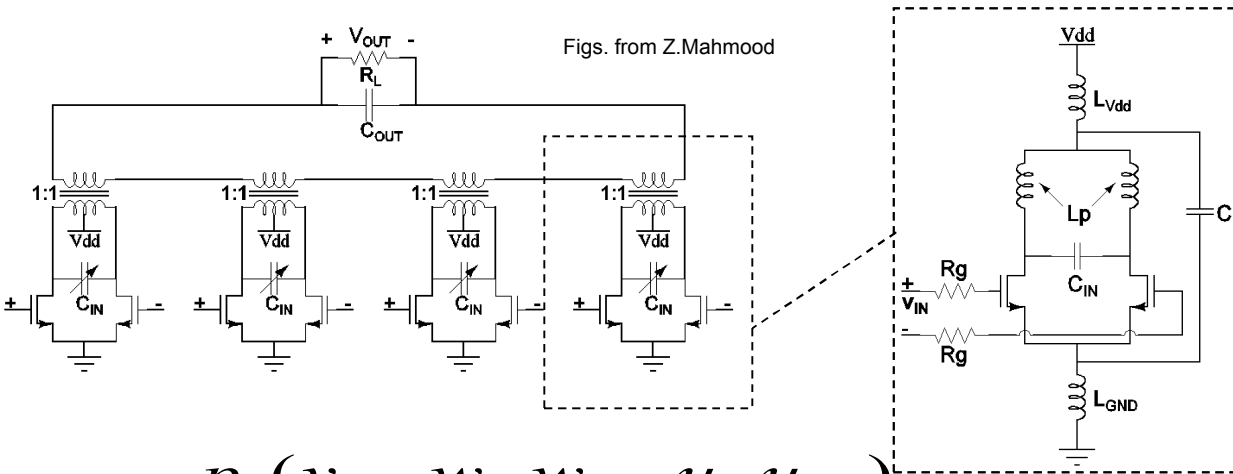
$$p = \alpha_0 + \alpha_1 u_t + \alpha_2 y_{t-1} + \dots + u_t^3$$

$$\underbrace{\alpha_{k,0} + \alpha_{k,1} \lambda + \alpha_{k,2} \lambda^2}_{\text{Coefficients of } \lambda}$$

Tested on 140 random pairs $\{\lambda_j, u_j(t)\}$

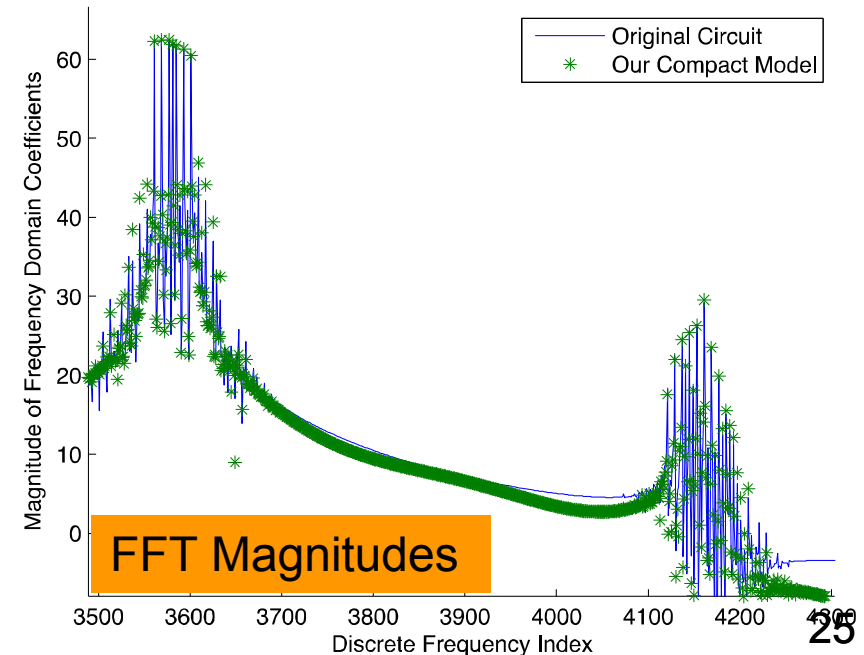
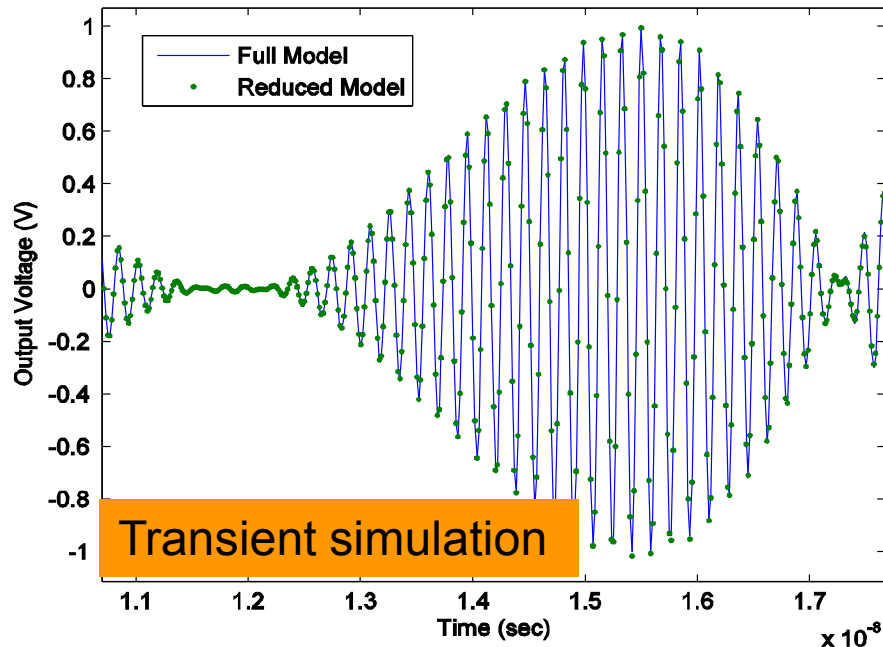


Example: Performance Analysis - Power Amplifier (1)

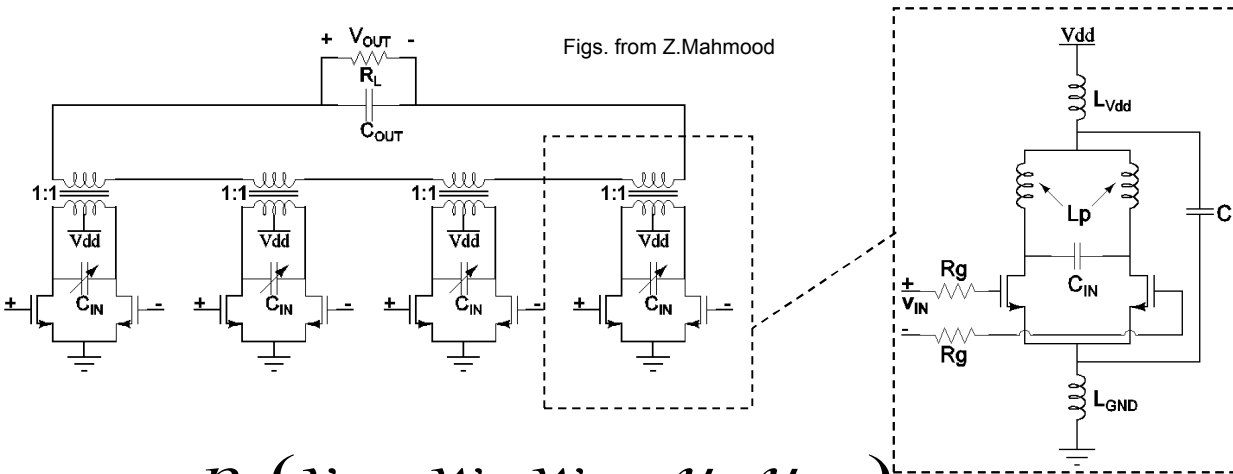


- Distributed power amplifier
- Identified rational model with feedback
- Fit also to LTI transformation of input
- 106 coefficients describing model

$$y_t = \frac{p_3(y_{t-1}, w_t, w_{t-1}, u_t, u_{t-1})}{q_4(y_{t-1}, w_t, w_{t-1}, u_t, u_{t-1})}$$

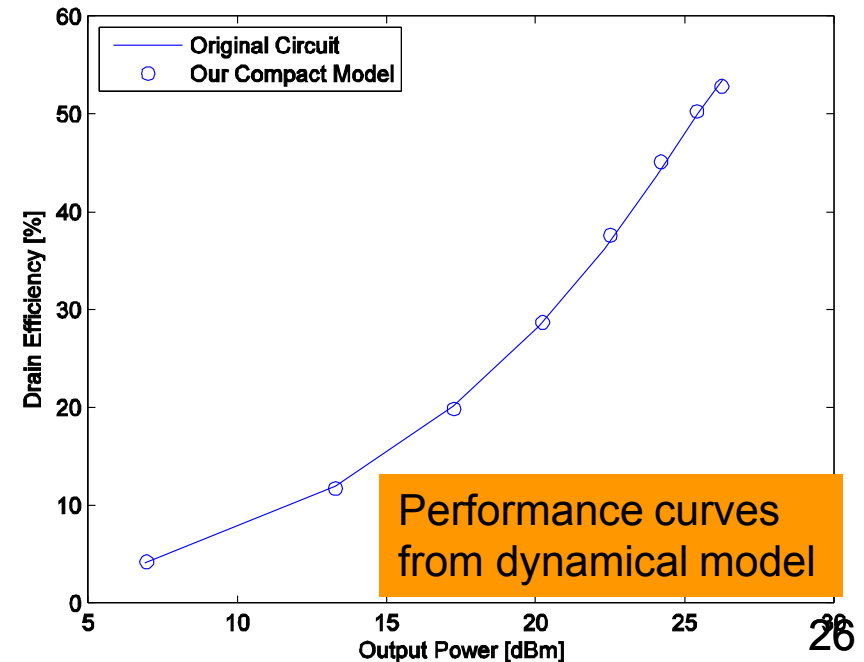
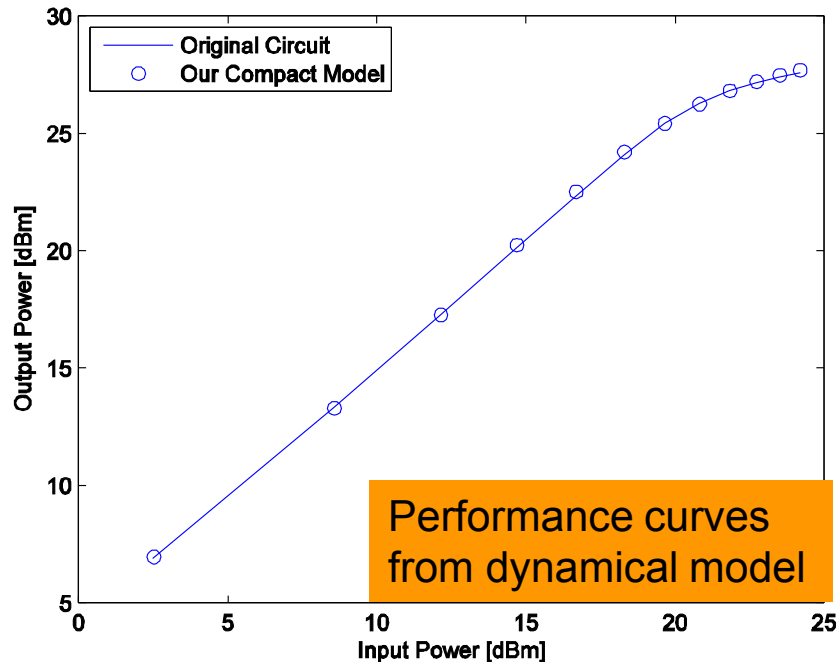


Example: Performance Analysis - Power Amplifier (2)

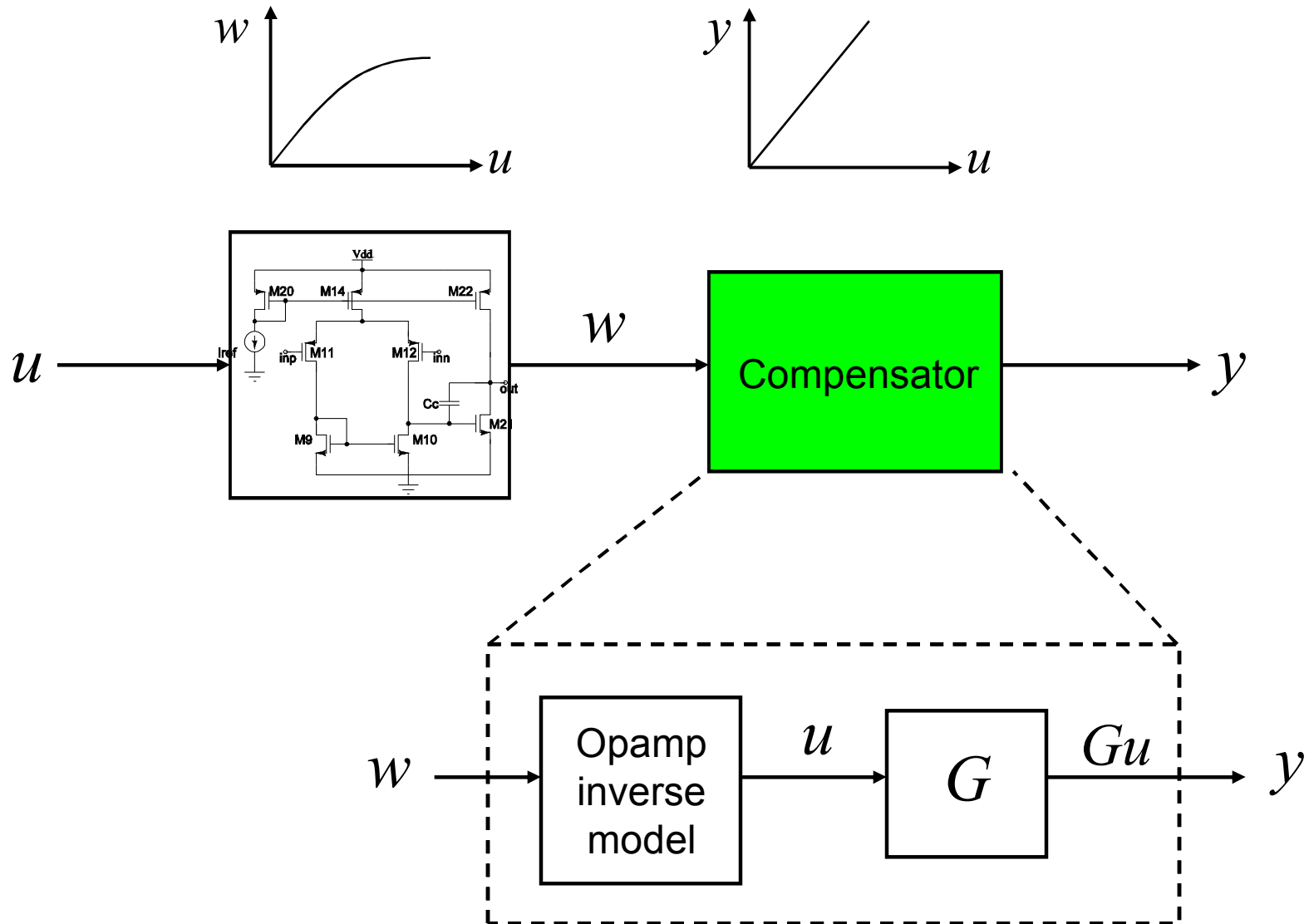


- Distributed power amplifier
- Identified rational model with feedback
- Fit also to LTI transformation of input
- 106 coefficients describing model

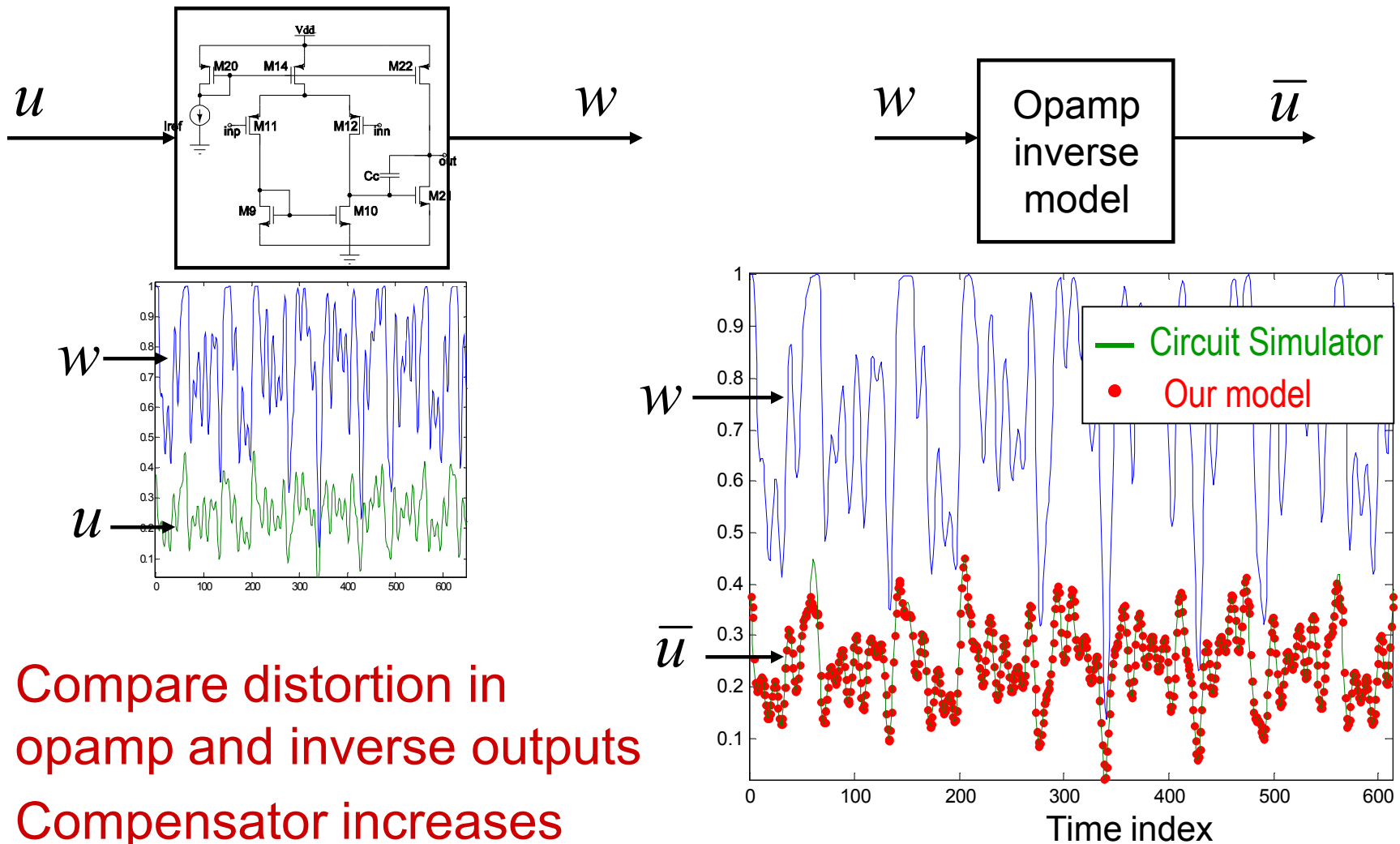
$$y_t = \frac{p_3(y_{t-1}, w_t, w_{t-1}, u_t, u_{t-1})}{q_4(y_{t-1}, w_t, w_{t-1}, u_t, u_{t-1})}$$



Example: Design - Opamp Compensator (1)

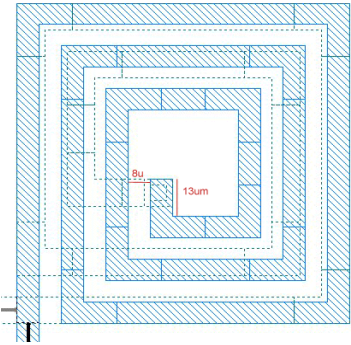


Example: Design - Opamp Compensator (2)



- Compare distortion in opamp and inverse outputs
- Compensator increases SDR by $\sim 10\text{dB}$

Example: Traditional LMOR Application - Inductor



Continuous Time, Linear Model

$$E\dot{x} = Ax + Bu$$

Simulate
original
system



$\{x, u\}$

Project
data

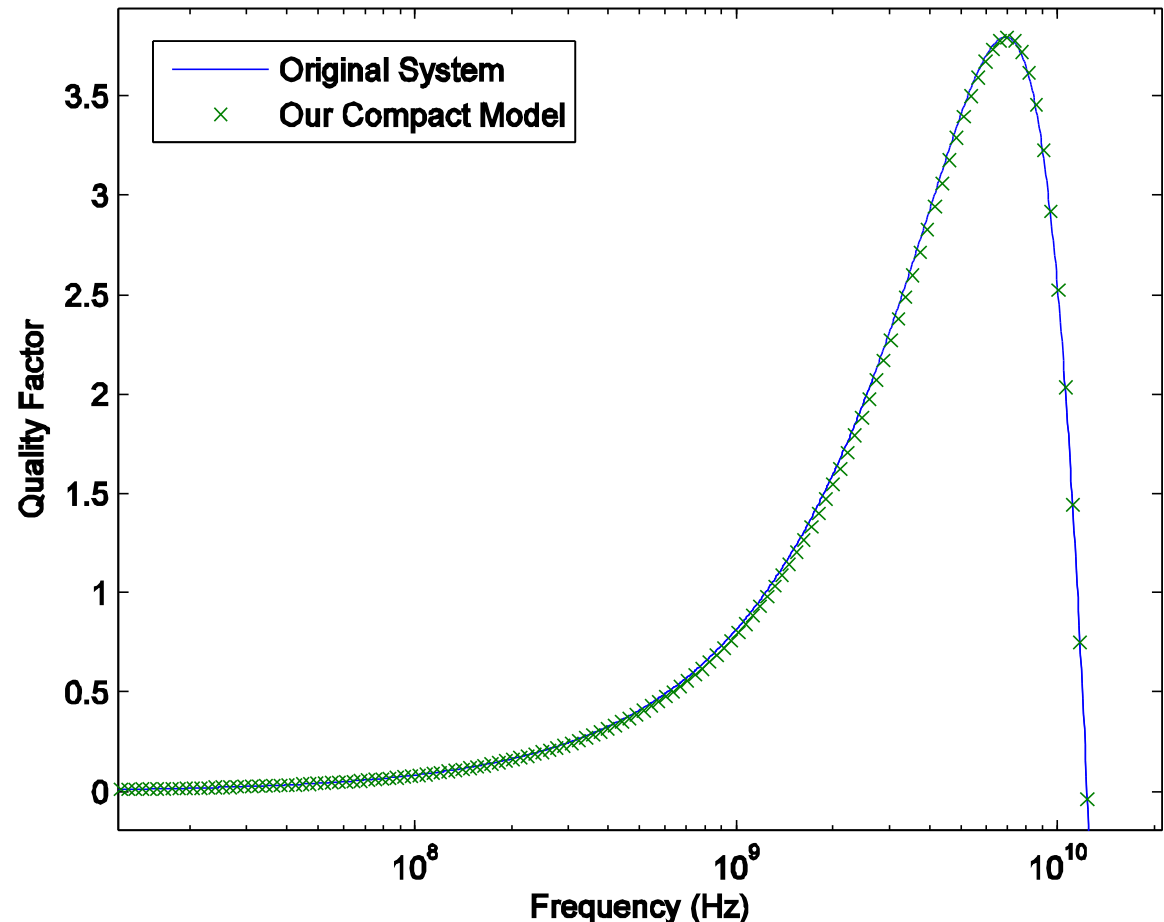


$\{\hat{x}, u\}$

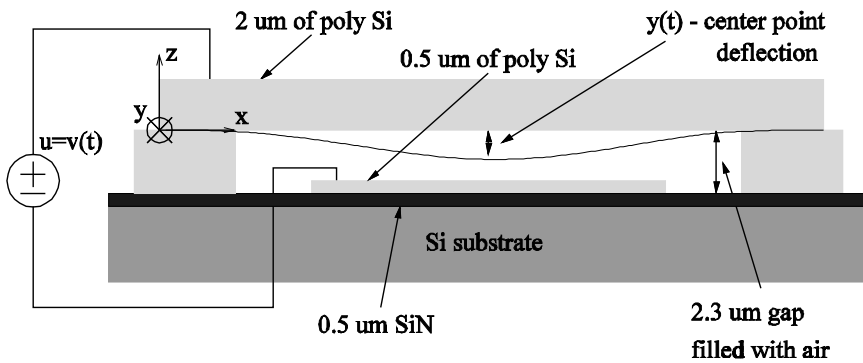
Identify
model



$$\hat{E}\dot{\hat{x}} = \hat{A}\hat{x} + \hat{B}u$$



Example: Traditional NLMOR Application - MEMS



400th order original model
4th order CT compact model
7th degree polynomial
52 coefficients
40X speedup for Transient

$$\dot{x} = f(x, u)$$

Simulate
original
system

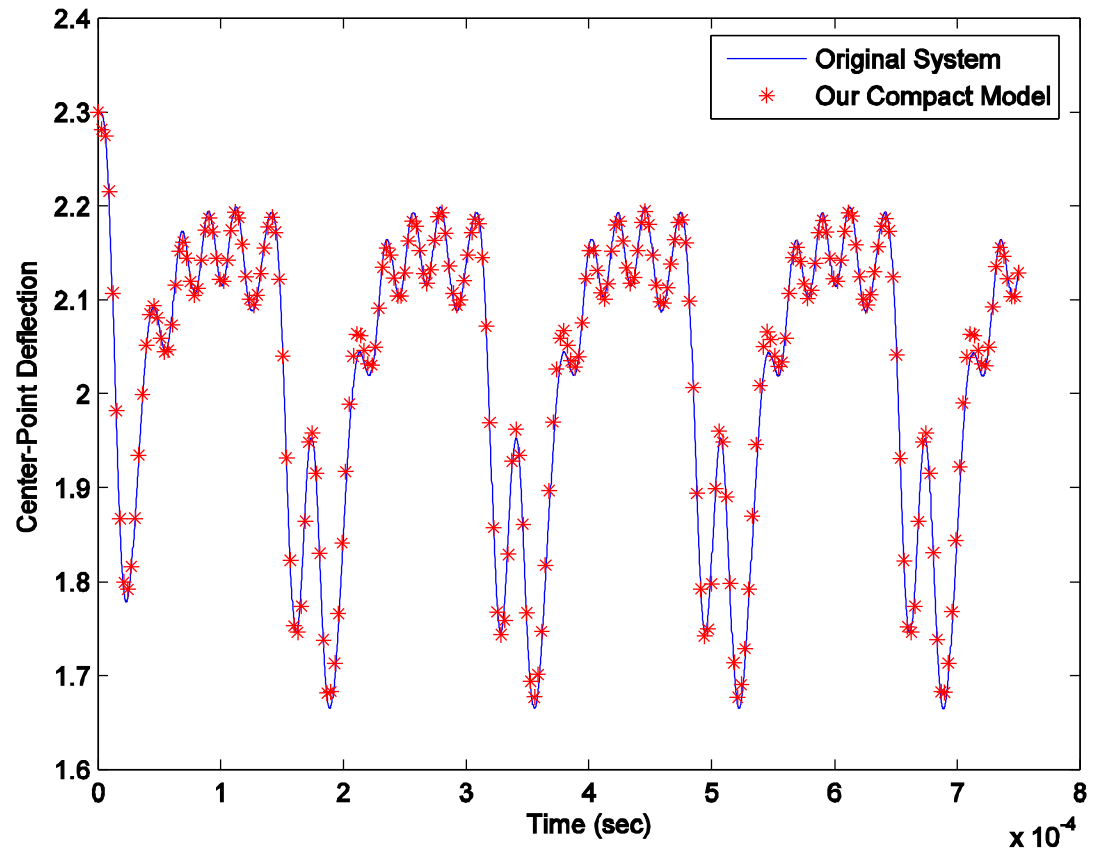
$$\{x, u\}$$

Project
data

$$\{\hat{x}, u\}$$

Identify
model

$$\dot{\hat{x}} = \hat{f}(\hat{x}, u)$$



Conclusion

- Accelerated simulation using compact models
- Need techniques guaranteeing robust models
- Encounter all types of systems in analog applications
 - Unstructured Linear, Nonlinear, Data samples
- SYSID approach based on semidefinite optimization

www.bnbond.com

Papers, codes, slides