

LA-UR- 11-01464

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Title: Metamaterial Based Terahertz Emitter

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Intended for: 2011 IC Postdoctoral Fellowship Research Colloquium



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Abstract for “Metamaterial Based Terahertz Emitter” presentation for 2011 IC Postdoctoral Research Fellowship Colloquium, 4-7 April 2011, Tysons Corner VA

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The objective of this research is the development of a metamaterial based terahertz (THz) emitter capable of operating at room temperature with only an applied DC bias voltage as power source. The current THz spectrum is underutilized and many applications exist for a compact metamaterial THz emitter (metamitter.) The concept is to integrate a negative differential resistance component into a split ring resonator in order to drive current and subsequent radiation. An ordered array of these gain-loaded elements will increase total emitted power. The present state of development of the metamitter device has encountered a technical hurdle with the micro-fabrication of the air-bridge connector for the resonant tunneling diode gain component. Concurrent research utilizing GHz scale analogs and passive THz frequency metamaterial arrays has been pursued in support of the metamitter design process. The GHz scale analogs have shown bias leads and frequency mixing to impact intended performance. The passive THz metamaterial response has shown that inter-element coupling greatly impacts the frequency and bandwidth of the measured resonances.

Presidential Early Career Program

Metamaterial Based Terahertz Emitter

M. T. Reiten

J. O'Hara, D. Roy Chowdhury, L. Earley, A. Azad, and A. Taylor

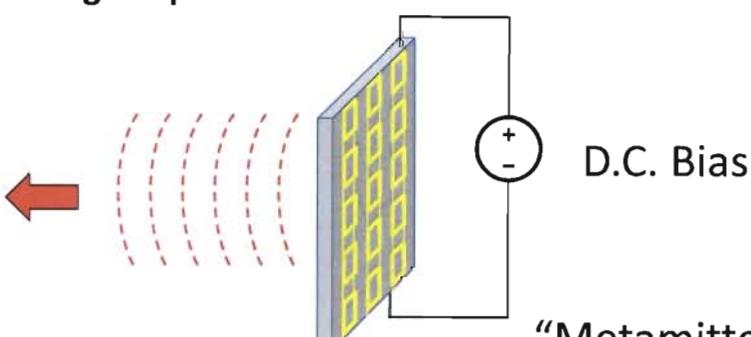
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Objective

Design, simulate, fabricate, and characterize active metamaterial based THz emitter (*metamitter*) capable of operating at room temperature with only an applied DC bias voltage as power source.



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Motivation



- Current THz spectrum is underutilized due to scarcity of sources, detectors, and natural limitations.

Approach	Operational frequency (THz)	Typical power (W)	Advantages	Disadvantages
Optical heterodyne	0.1-3.0	10^{-8} - 10^{-5}	Relatively simple concept	Poor power
Quantum cascade laser	0.1-5	10^{-3}	Good power	Difficult fabrication. Needs cryogenic cooling
Gas laser	0.1-10	10^{-3} - 10^{-1}	High power	Large. Fixed operating frequencies
Tube sources	0.1-1.0	10^{-3} - 10^{-3}	Excellent power	Fabrication difficulty. Requires huge magnetic fields for high power
Free-electron laser	0.1-10	10^{-3} - 10^{-6}	Excellent power	Building-sized system
Solid-state	< 1	10^{-6} - 10^{-9}	Stable. Reliable. Compact	Power very low approaching 1 THz

- THz spectroscopy (and imaging) has been confined to the laboratory and/or controlled environments.



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Metamitter Applications



- Compact continuous wave THz emitter
 - Non-invasive imaging source
 - Targeted spectroscopy using tuned arrays for:
 - Security (explosive and chemical detection)
 - Pharmaceutical
 - Bio Applications
- Potential to lead to THz focal plane array



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Concept

- Integrate negative differential resistance component (RTD) into metamaterial element in order to drive current oscillation and subsequent radiation.
- Create ordered array of radiating elements to increase total emitted power.
- Potential to shape the radiation pattern by modifying the array.

Metamaterials

- **Metamaterials:** Engineered materials with derived from the composite properties of the structure and the surrounding material that can exhibit a tailored effective medium behavior. The Split Ring Resonator (SRR) can serve as a metamaterial element which has both electric and magnetic response.[1,2]

Potential of metamaterials: Negative Index Materials → Cloaking Screens, Perfect Lenses; Ability to tune parameters

[1] J.B. Pendry et al., "Extremely Low Frequency Plasmons in Metallic Mesostructures," *PRL* **76**, 4773 (1996)
 [2] J.B. Pendry et al., "Magnetism from conductors and enhanced nonlinear phenomena," *IEEE Trans. Microwave Tech.* **47**, 2075 (1999)

Metamitter Element

Tank circuit provides feedback

Gap 1

Gap 2

RTD

RTD Gain

Frequency

SRR Resonances

Air Bridge

Active Layers

InP Substrate

Mesa (Contact Layer)

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SRR

Split Ring Resonator (SRR)

- Asymmetry important for radiation coupling from fundamental resonance
 - Gap ratio
 - Gap placement
- Two gaps required for bias

30°

10°

Fedotov, et. al., PRL 99, 147401 (2007)

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Al-Naib, et. al., APL 94, 153505 (2009)

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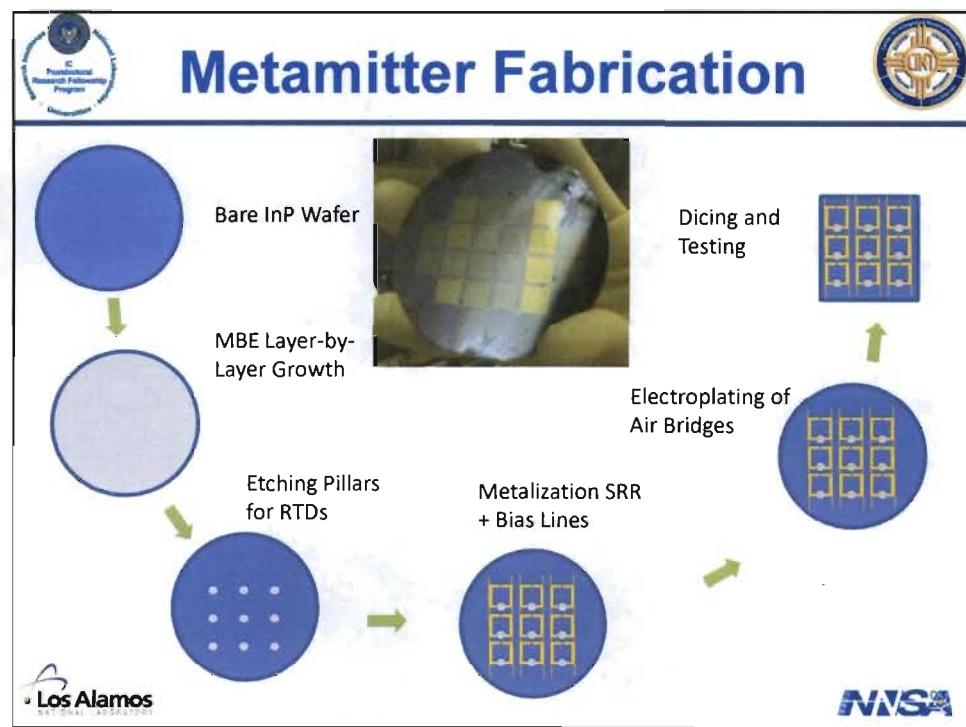
RTD

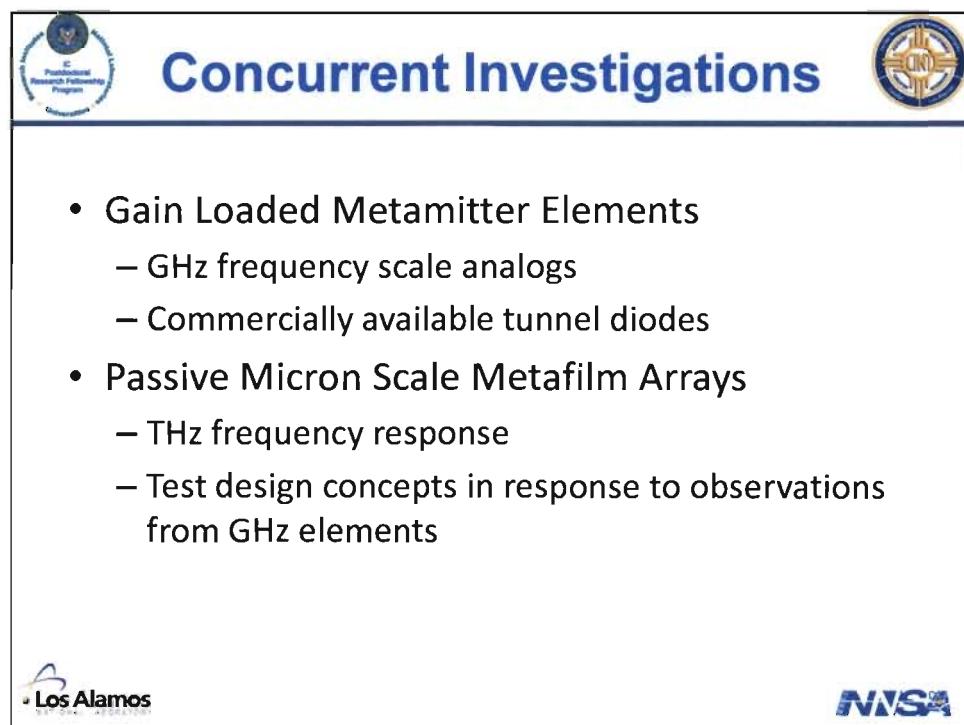
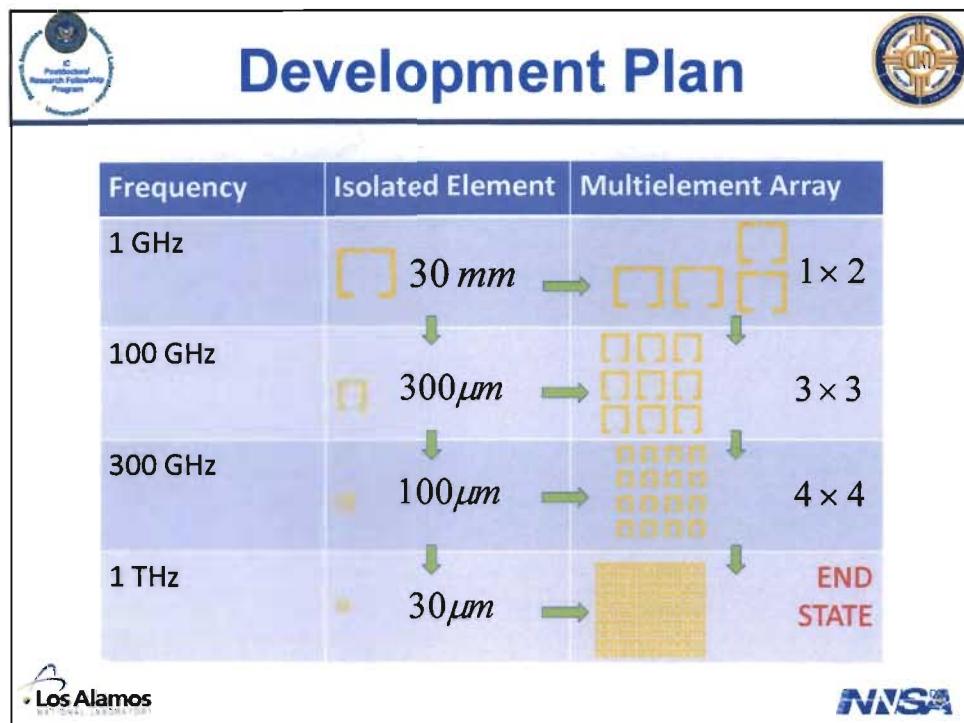
Resonant Tunneling Diode (RTD)

- Negative Conductance Device
- Resonant Tunneling
 - Two barriers leads to 100% transmission
 - High Frequency Response

The diagram shows a cross-section of an RTD structure. It consists of a central 'Metal Air Bridge' layer sandwiched between two 'InGaAs/Al As layers'. The structure is mounted on a substrate. Above the diagram is a graph of Current (I) versus Voltage (V). The current increases with voltage until it reaches a peak and then decreases, forming a negative conductance region. A vertical dashed line marks the DC bias voltage V_{DC} .

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 **GHz Metamitters** 

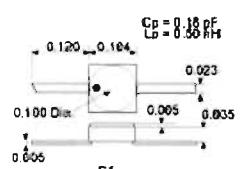




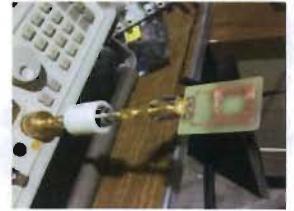
 

 **Tunnel Diodes** 

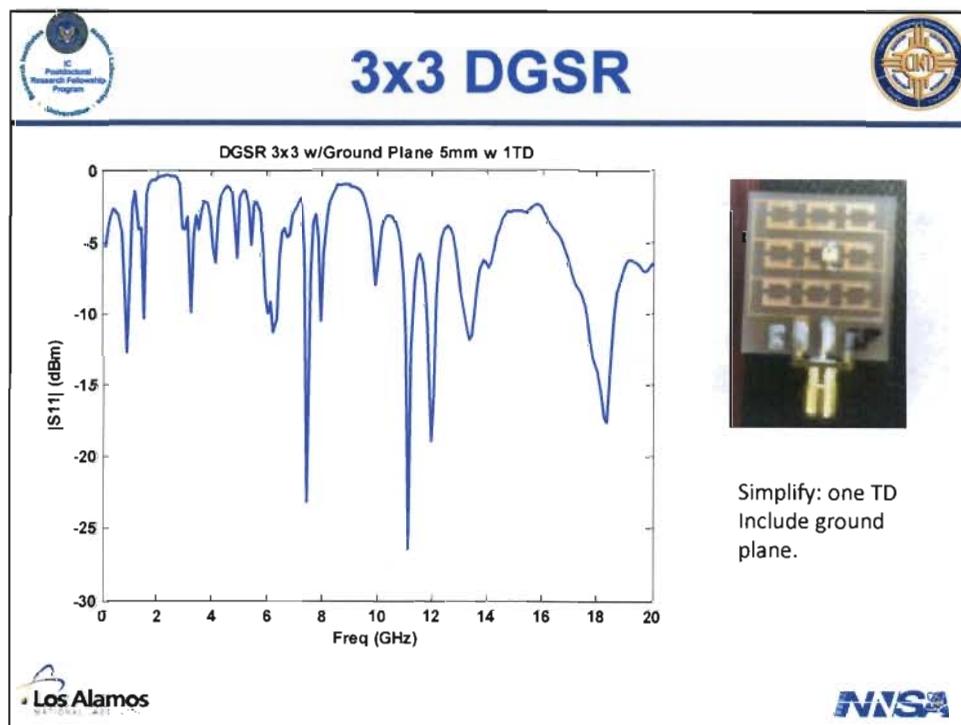
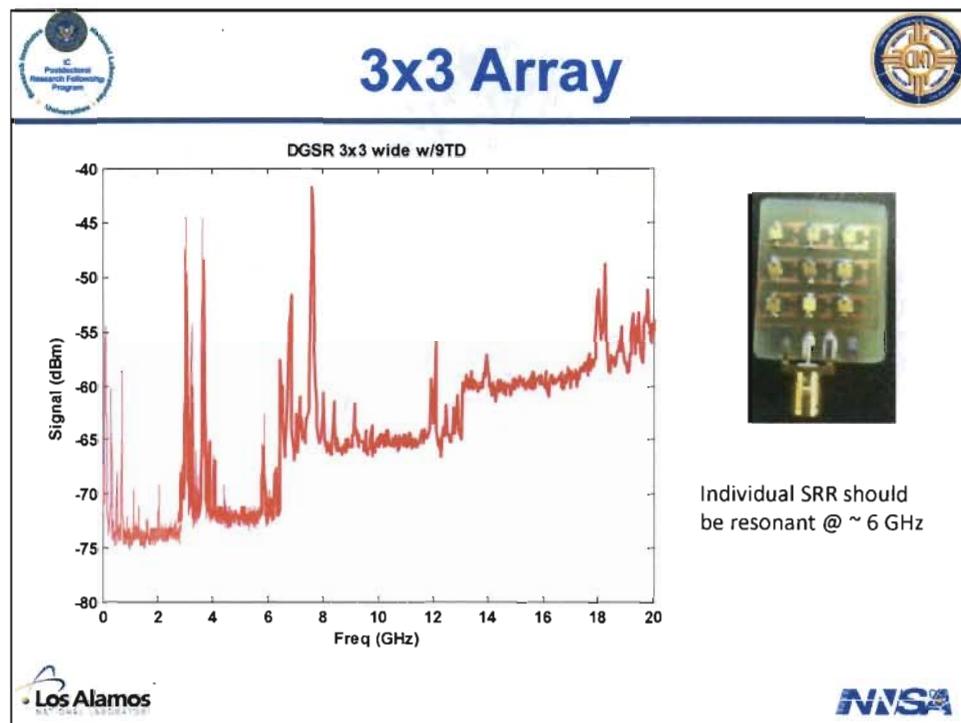
- Meant to be a GHz frequency stand in for THz RTDs
 - Exhibits same negative differential resistance
 - Low threshold voltage
- Can the nonlinearity of the TD account for all the frequency mixing behavior?

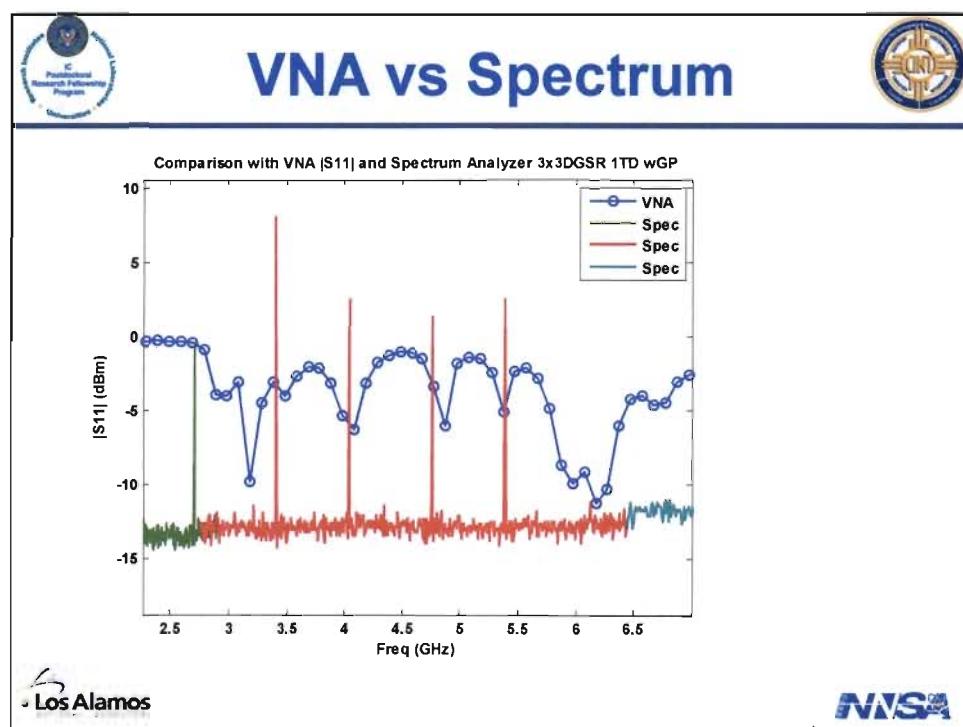
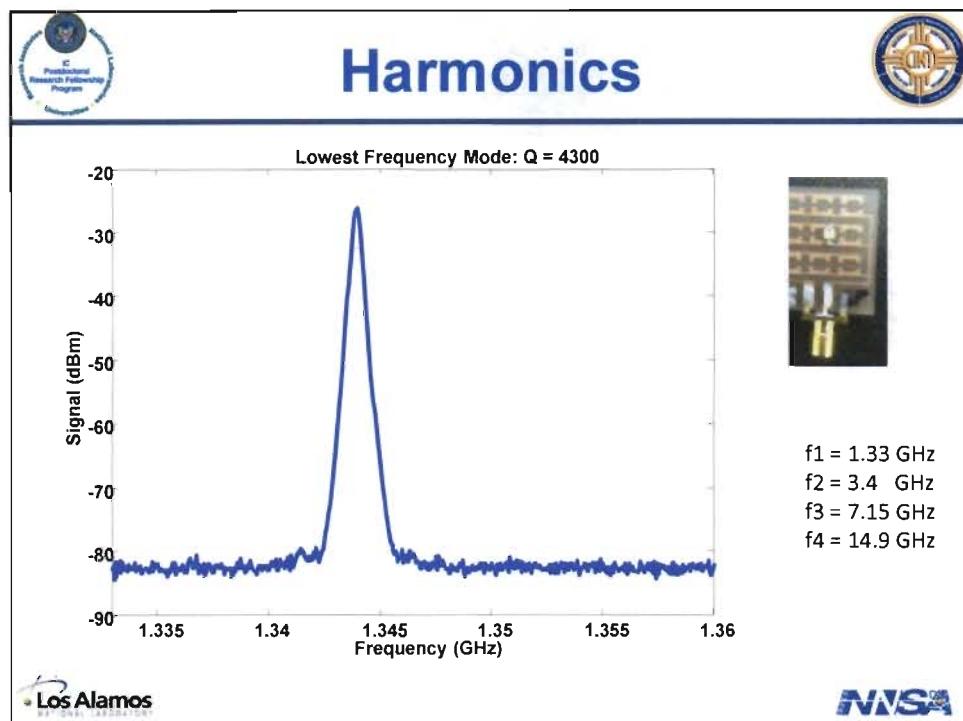


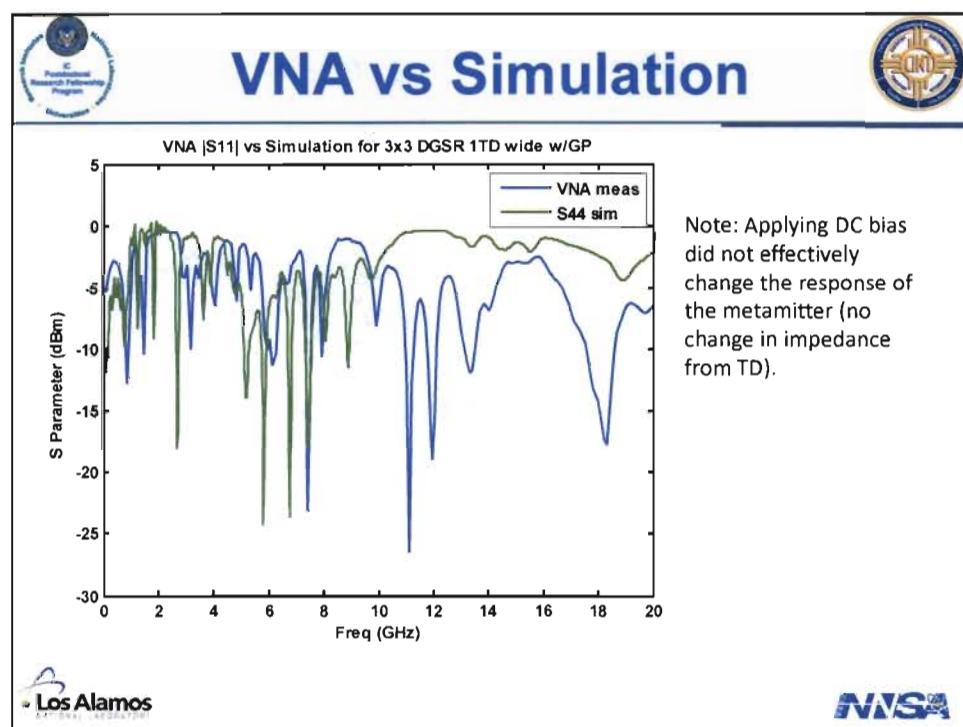
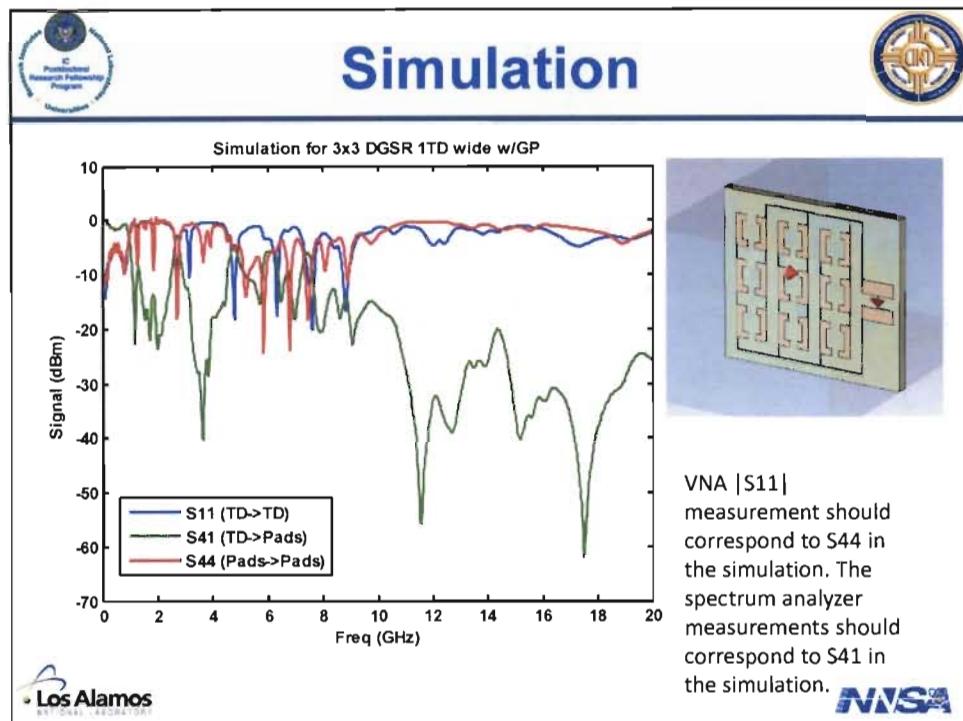
http://www.mpulsemw.com/Tunnel_Diode.htm

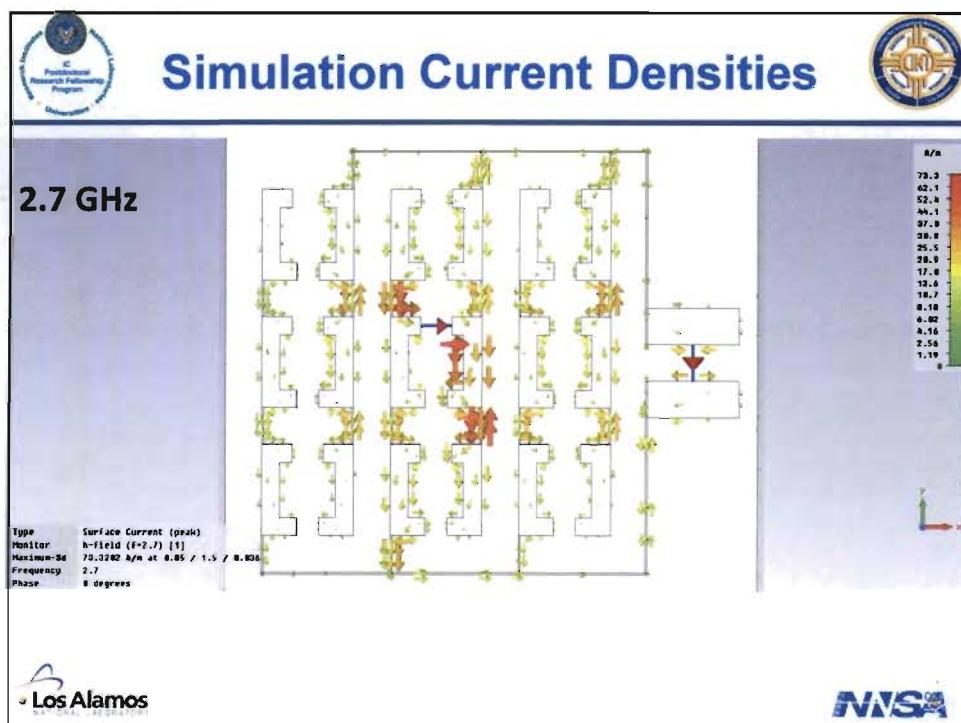
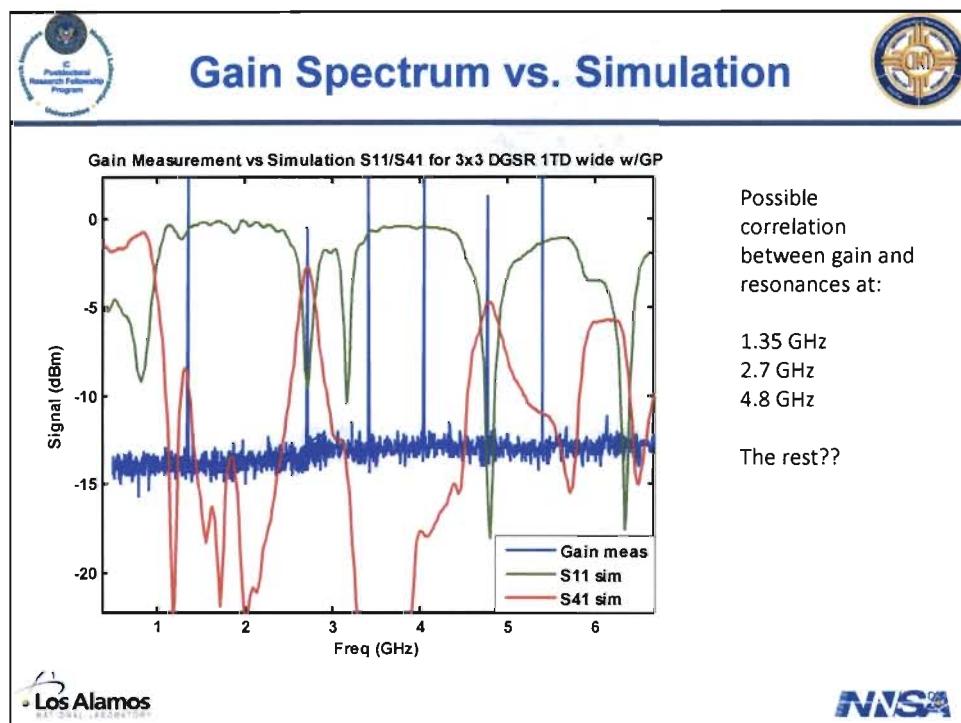


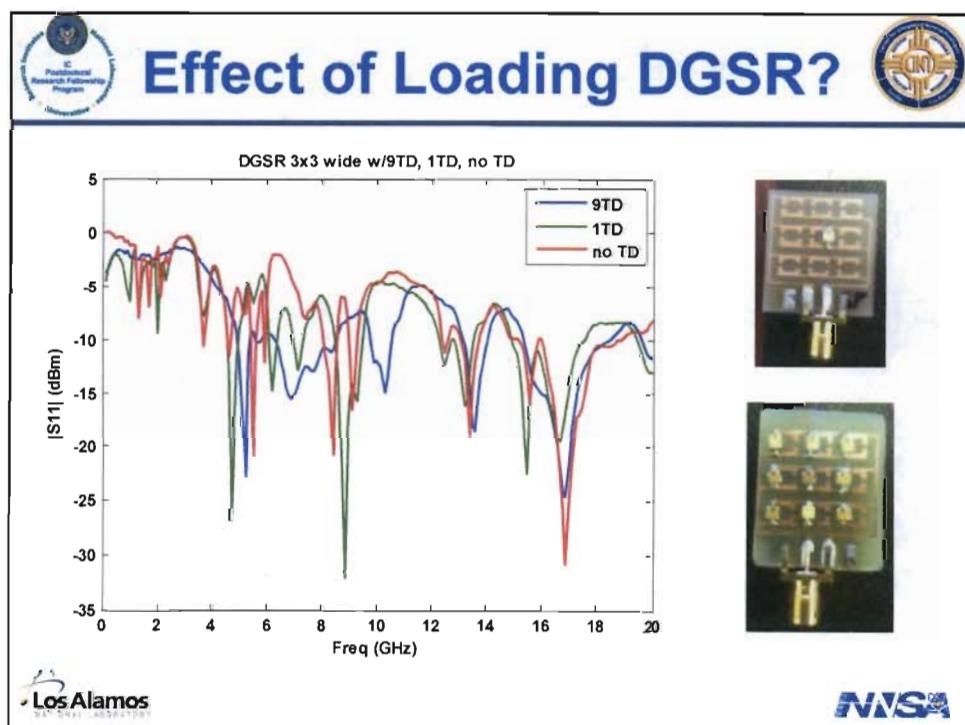
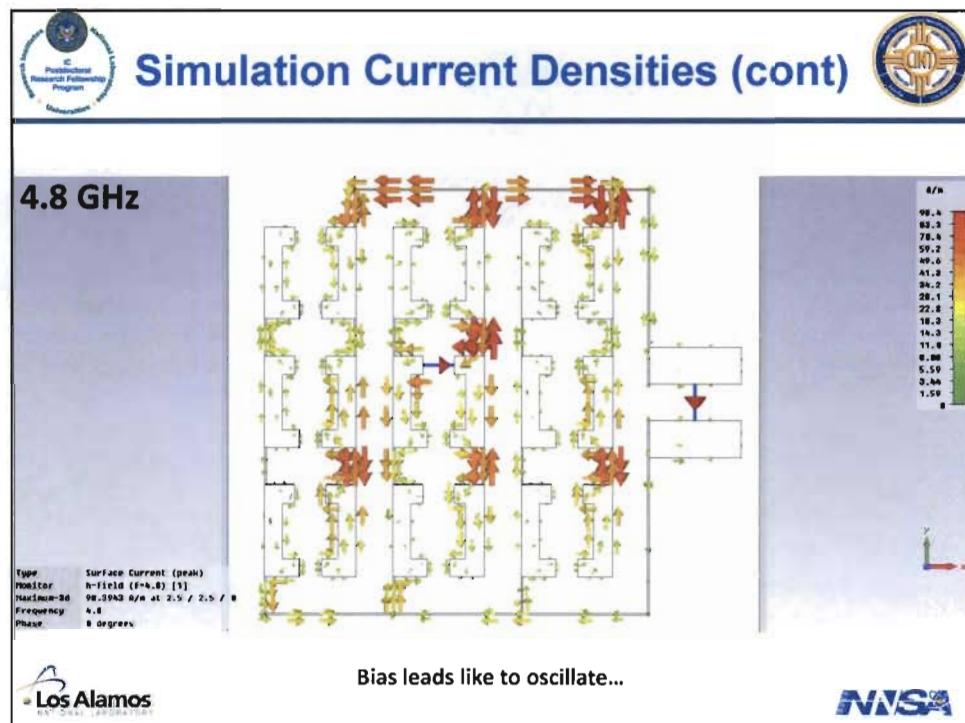
 

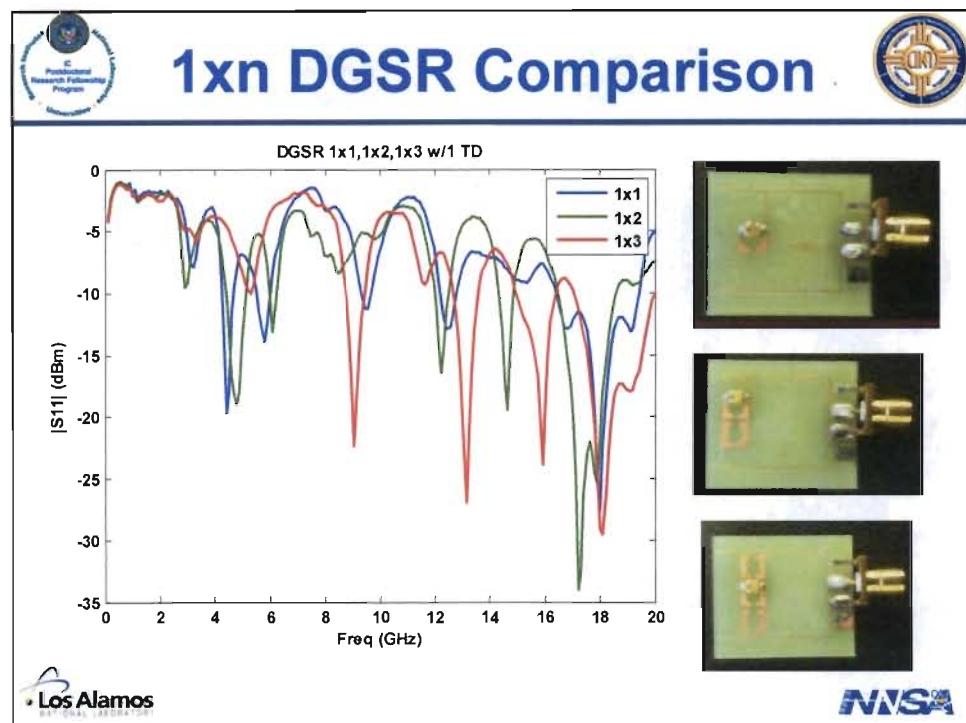










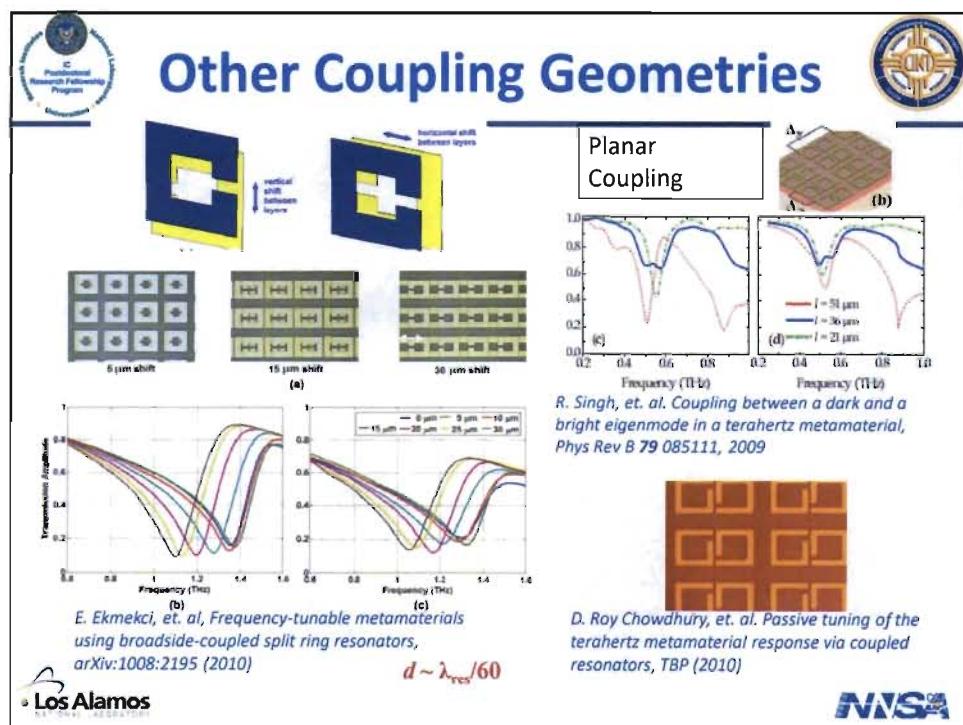
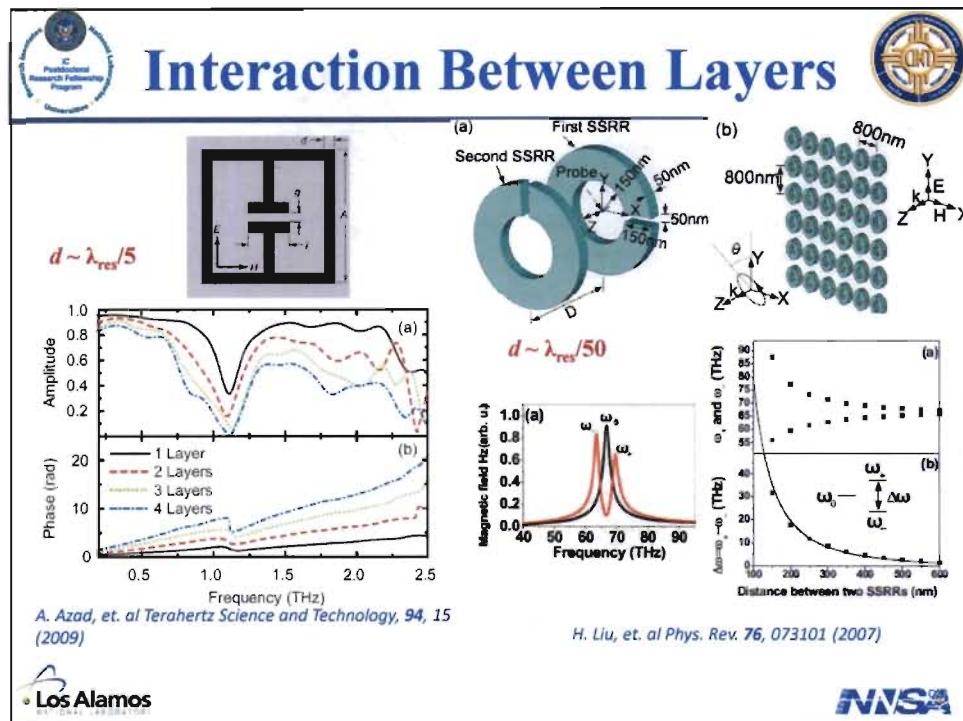


Identified Challenges

- DC Bias Lines?
 - Had wanted to avoid lumped elements for scaling
 - Introduction of lossy elements inconclusive
- Measurement
 - $|S11|$ misleading as designed
 - Coupling to free space may be best
- Coupled Oscillators
 - General result: n oscillators $\rightarrow n$ resonances (may be degenerate if weakly coupled)
- Multiple Frequency Components
 - Resonant Feedback to Stabilize Oscillators?

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Previously Observed Resonance Shifts

The Lagrangian of an isolated SRR

$$\mathcal{L} = \frac{LQ^2}{2} - \frac{Q^2}{2C} \quad \omega_0 = \sqrt{1/LC}$$

The Lagrangian of coupled SRRs

$$\mathcal{L} = \frac{LQ_1^2}{2} - \frac{Q_1^2}{2C} + \frac{LQ_2^2}{2} - \frac{Q_2^2}{2C} + M_H Q_1 Q_2 + M_E \omega_0^2 Q_1 Q_2 \cos \varphi$$

Solving Euler-Lagrange equation

$$\frac{d}{dt} \left(\frac{\partial \mathcal{L}}{\partial Q_i} \right) - \frac{\partial \mathcal{L}}{\partial Q_i} = 0, \quad i = 1, 2$$

$\omega_{\pm} = \omega_0 \sqrt{1 \mp \frac{M_E \cos \varphi}{L}}$ Where did C go in this process?
Wrapped up in $\omega_o = 1/\sqrt{LC}$

N. Liu, et. al. Nature Photonics 3, 157 (2009)

$d \sim \lambda_{\text{res}}/500$?

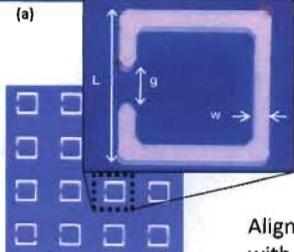
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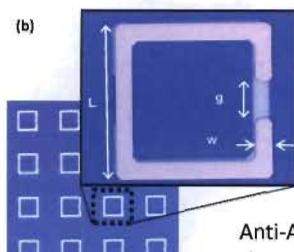
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Our Experiment

Sample Fabrication

• Standard photolithography
 • Substrate Si-GaAs
 • Spacing Layer d Polyimide
 • SRR: 200 nm Au
 • Unit Cell: 98 μm
 • Thickness: $d = \sim 1$ and 4 μm

(a) 
Aligned SRRs with $d=1 \mu\text{m}$

(b) 
Anti-Aligned SRR with $d=1 \mu\text{m}$

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Simulation for Bilayer Arrays Varying d

Lower Resonances ω_- and Isolated SRR resonances

Polyimide Layer	1 μm (sim)	4 μm (sim)
Aligned	0.34	0.33
Anti-aligned	0.15	0.25
SRR Top Only	0.42	0.53
SRR Bottom Only	0.35	0.34

CST Microwave Studio 2009 Time Domain Solver (FDTD) used to simulate the THz TDS experiment.

Simulation of Transmission through Bilayer SRR Array

Consequence of inhomogeneity is the SRRs on top and bottom will experience different dielectric environments and will have different uncoupled resonances.

SRR Top and Bottom Resonances at 4 μm

SRR Top and Bottom Resonances at 1 μm

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Eigenfrequencies for Inhomogenous Bilayer

$$\Gamma = \frac{L_1}{2} (\dot{Q}_1^2 + \omega_1^2 Q_1^2) + \frac{L_2}{2} (\dot{Q}_2^2 + \omega_2^2 Q_2^2) + M_H \dot{Q}_1 \dot{Q}_2 \pm M_E Q_1 Q_2$$

Solving Euler-Lagrange eqn..., realizing that $L_1 = L_2$...assuming a harmonic oscillator solution $Q_n = B_n e^{i\omega t}$ and then taking the determinant of the resulting coupled differential equations yields:

$$(\omega^2 - \omega_1^2)(\omega^2 - \omega_2^2) - \left(\frac{M_H}{L} \omega^2 \pm \frac{M_E}{L} \right)^2 = 0$$

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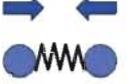
Elementary Coupled Oscillators

- General result: n coupled oscillators = n resonant frequencies (may be degeneracies)
- Consider simplest case of two coupled “mass on a spring” oscillators attached to a fixed surface with a spring constant of κ_{self} (not pictured) and coupled with spring constant of $\kappa_{coupling}$



Symmetric
(lower energy)

$$\omega_- = \sqrt{\frac{\kappa_{self}}{m}}$$



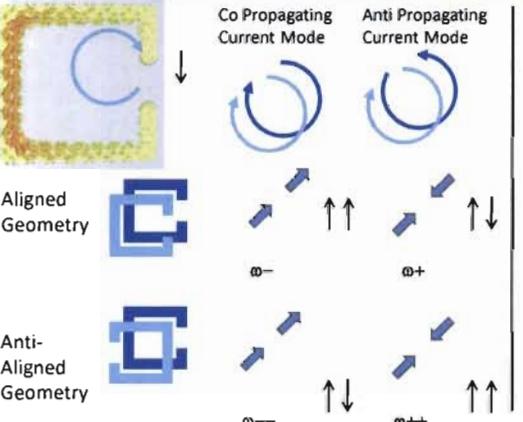
Anti-symmetric
(higher energy)

$$\omega_+ = \sqrt{\frac{\kappa_{self} + 2\kappa_{coupling}}{m}}$$

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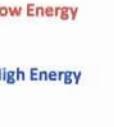
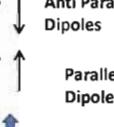
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Current Modes and Coupling Mechanisms



Aligned Geometry

Anti-Aligned Geometry

<p>Co Propagating Current Mode</p> 	<p>Anti Propagating Current Mode</p> 
<p>Low Energy</p> 	
<p>High Energy</p> 	
<p>Low Energy</p> 	
<p>High Energy</p> 	

With a single coupling mechanism it is relatively easy to identify the “symmetric” and “anti-symmetric” modes. Two competing coupling mechanisms complicate the process.

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8F THz Time Domain Spectroscopy System

• Photoconductive Switch System
 • Sample placed at frequency independent focus
 • 2:3 Imaging at focus
 • Beam illuminated ~ 2500 SRR pairs
 • Etalon pulses limit measurement window to 18.33 ps (Sample substrate & transmitter chip GaAs with thickness of 500 μ m)

THz RCV

Sample

THz TRX

BS

Ti:Sapphire Oscillator

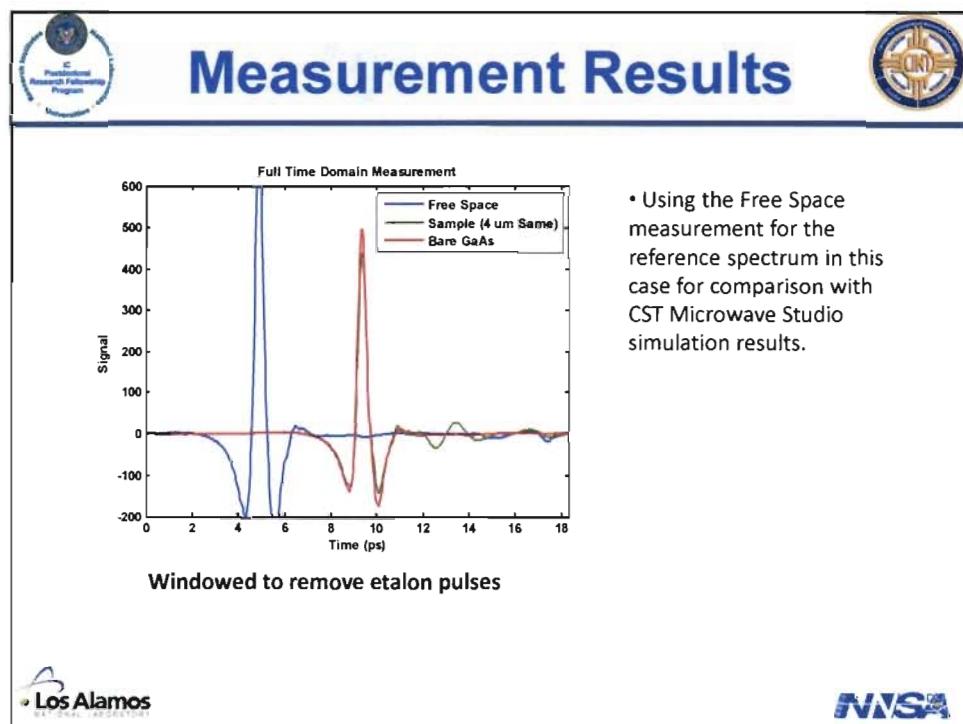
Delay Line

THz RCV

• Photoconductive Switch System
 • Sample placed at frequency independent focus
 • 2:3 Imaging at focus
 • Beam illuminated ~ 2500 SRR pairs
 • Etalon pulses limit measurement window to 18.33 ps (Sample substrate & transmitter chip GaAs with thickness of 500 μ m)

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Time Windowing Effect on Spectra

0.5 THz CW Signal Truncated before FFT and Padded to 1k pts

Perfect Sine Wave at 0.5 THz
 $\Delta t = 0.01$ ps
 $N = 1000$ pts

For low frequency resonances, few cycles will be captured in a finite time window.

Zero padding doesn't add signal, but it does make the result look more continuous.

May add artifacts in analysis.

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Sim. vs. Meas. for Aligned SRR

Transmission Through Aligned SRR Bilayer Sample

Transmission

Frequency (THz)

$d = 1 \mu\text{m}$

$d = 4 \mu\text{m}$

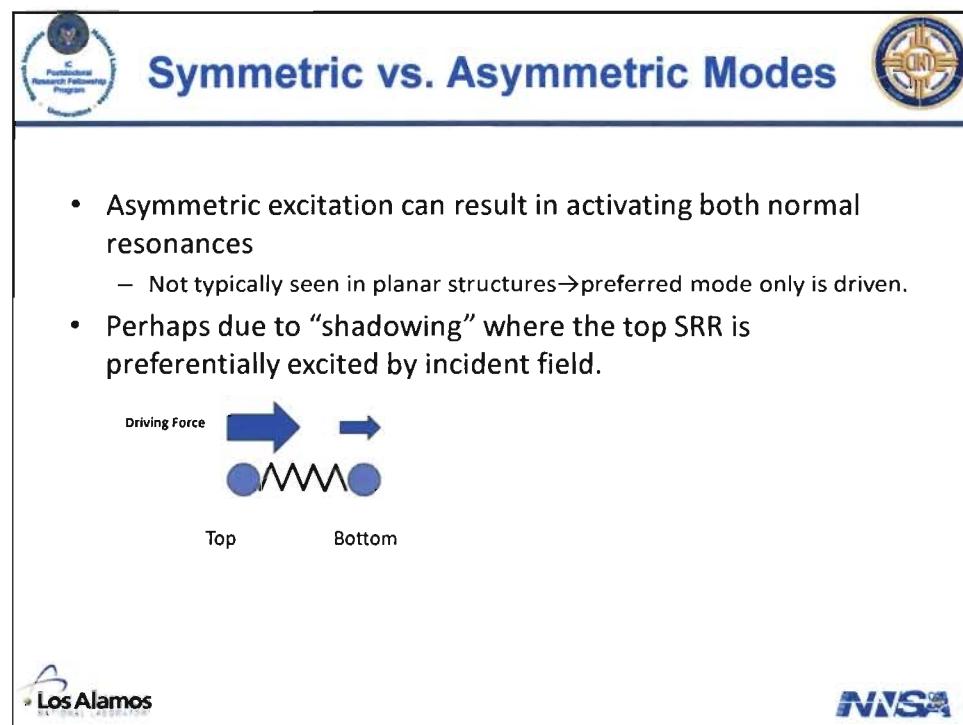
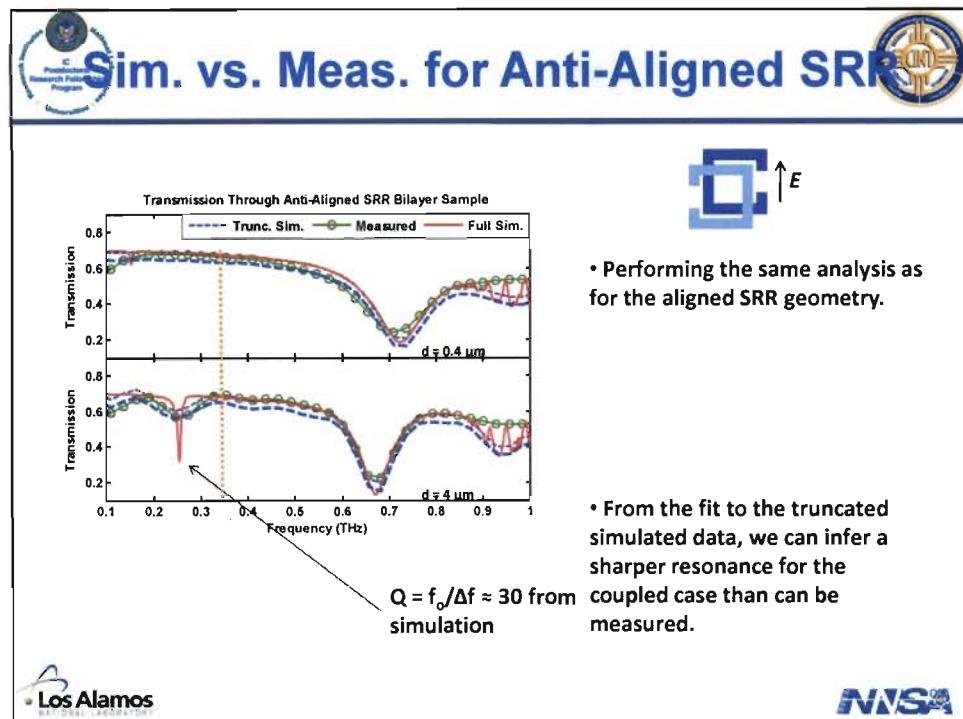
$Q = f_o/\Delta f \approx 12$ from simulation

E

- Measurement data zero padded to 36.66 ps (2x time window)
- $|S_{21}|$ response adjusted for additional Fresnel losses
- Structure beyond ~0.85 THz likely due to higher order dipoles and interaction with boundary conditions/waveguide modes.
- Truncated time domain analyzed like measurement (18.33 ps window, 2x zero padded)

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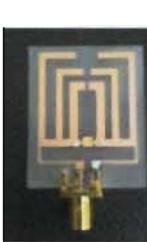
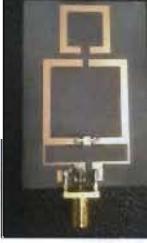


Coupled Oscillator Research

- Coupled oscillators follow fundamental physics. Identifying and isolating the coupling mechanisms can be a challenge in order to make use of the fundamental physics to gain further insight.
- Applied: Have method to enhance Q by coupling resonators which may lead to more efficient oscillators and have red shifted the resonance which may lead to more effective electrically small antennas.
- Experimental: In order to resolve high Q resonances, the time window available must be increased for measurements.
- Future: Working toward extracting the relative strength of the electric and magnetic dipole coupling as a function of d for various geometries.

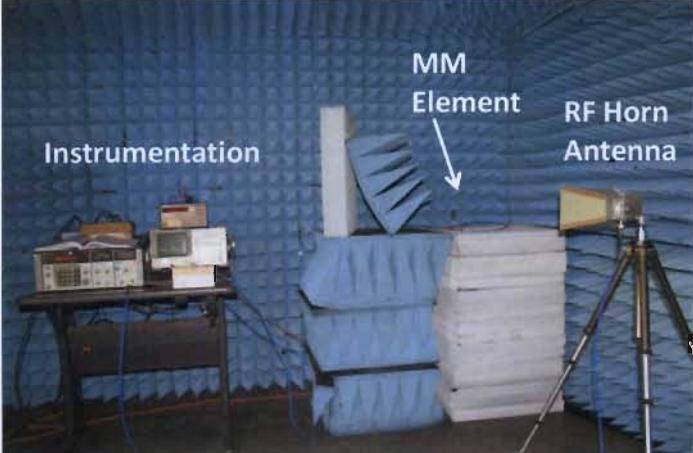
GHz Metamitter Elements

	DGSR (Large) 20 mm x 20 mm 2 mm gap 5 mm width $f_0 \approx 2.4$ GHz		DGMR 25 mm x 25 mm outer 16 mm x 19 mm mid 9 mm x 12 mm inner 6/2/1 mm gaps 1.5/1.5/ 2 mm widths $f_0 \approx 1$ GHz (approx)
	DGDR 25 mm x 25 mm outer 16.3 mm x 16.3 mm inner 1/1.75 mm gaps 2 mm width $f_0 \approx 1$ GHz		DGSR+SGSR 25 mm x 25 mm DGSR 15 mm x 15 mm SGSR 2/4 mm gaps 2 mm width $f_0 \approx 1$ GHz

Experimental Set Up

External free space RF excitation experiment



Instrumentation

MM Element

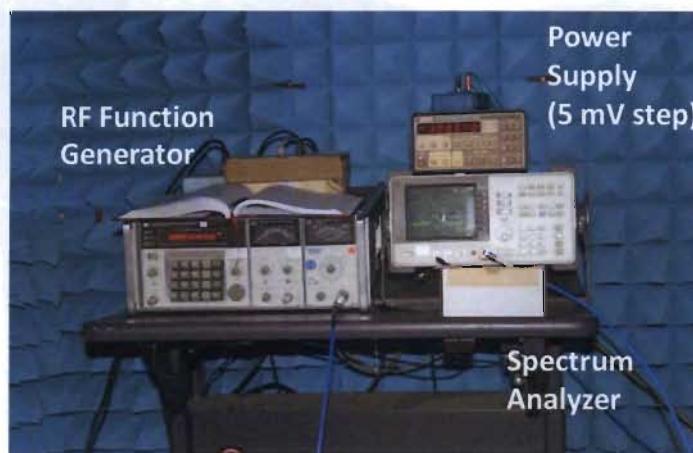
RF Horn Antenna

All external RF measurements were taken with magnetic excitation of the metamaterial element. No contribution expected from the

Los Alamos (Photo Courtesy Larry Earley)

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Instrumentation



RF Function Generator

Power Supply (5 mV step)

Spectrum Analyzer

Self oscillation measured with power supply and output on the spectrum analyzer.

Los Alamos (Photo Courtesy Larry Earley)

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DGSR Metamitter

Self Oscillation of Single DGSR (large) with Bias Voltage

Note onset of sidebands with increasing bias.

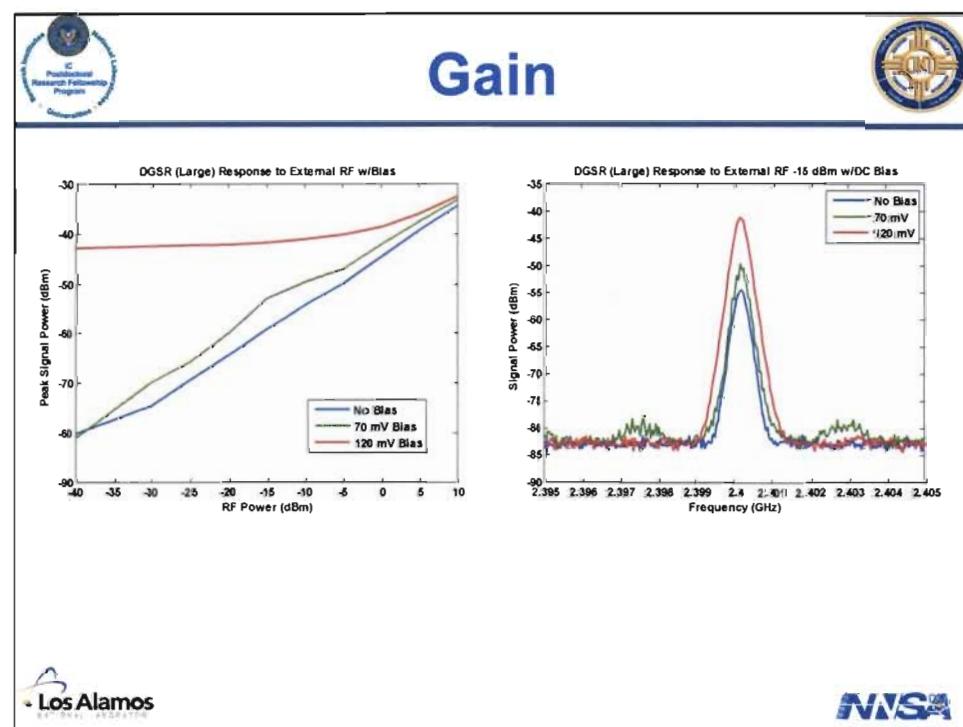
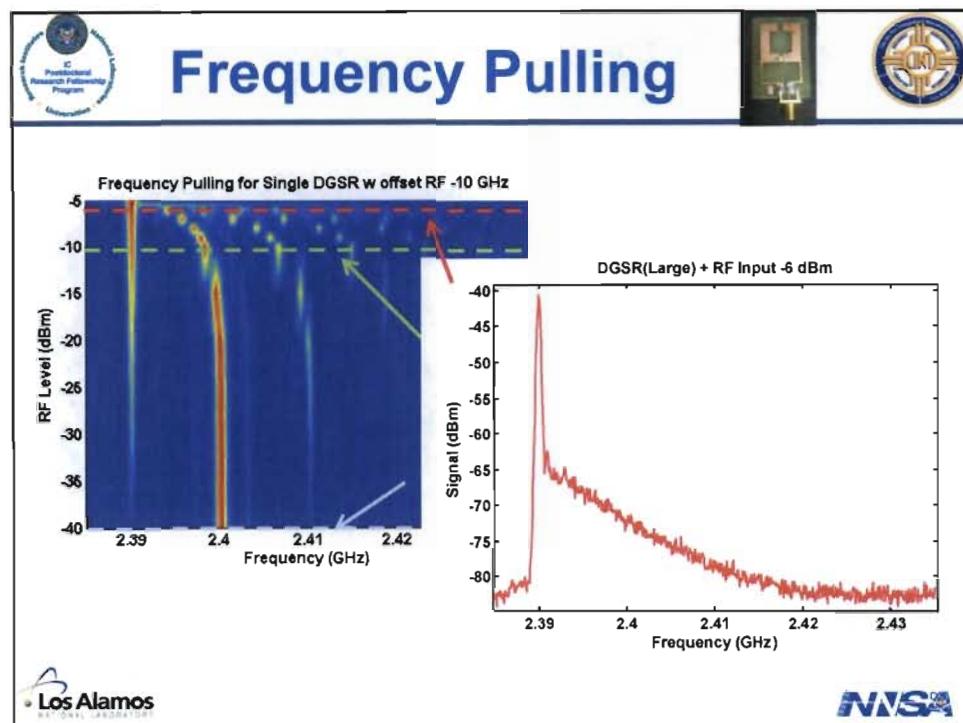
Frequency shifting is most likely due to variation in the gain (slope = dl/dV) of the negative differential resistance.

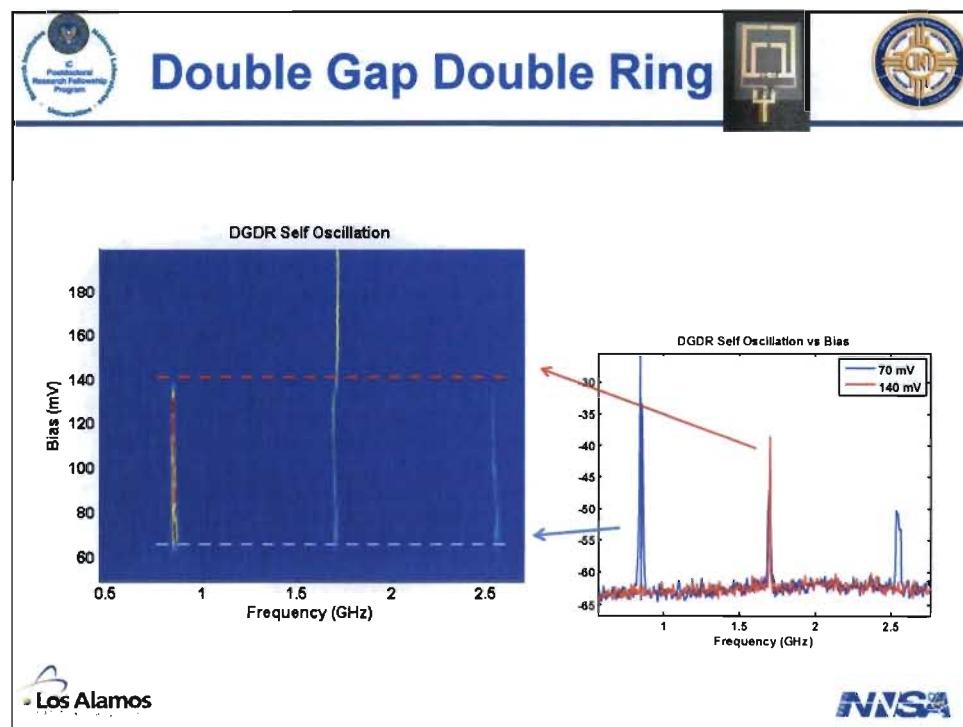
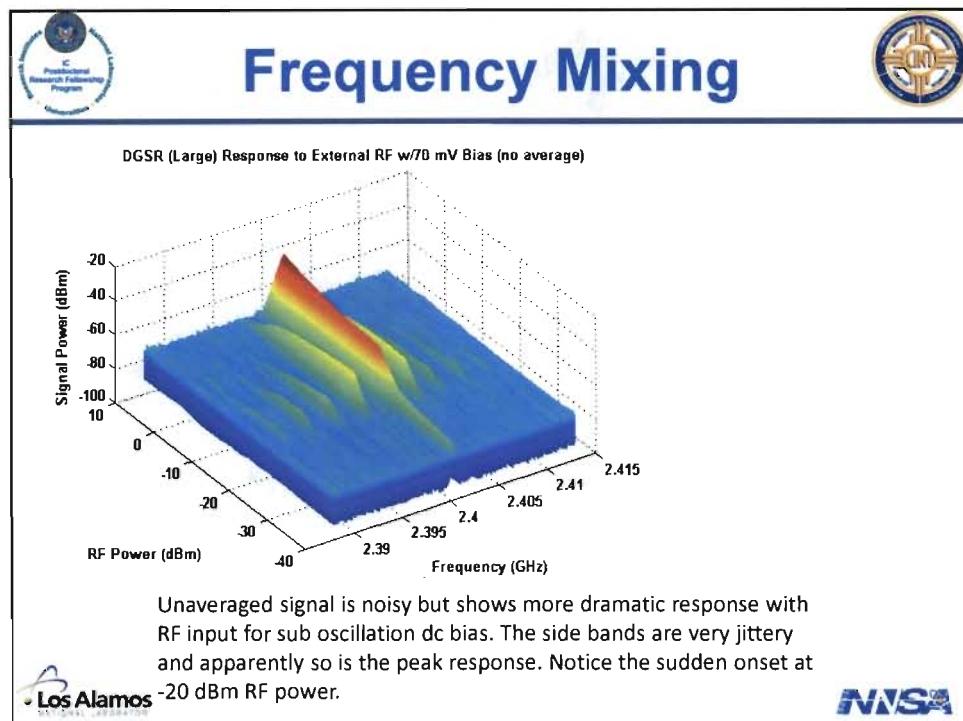
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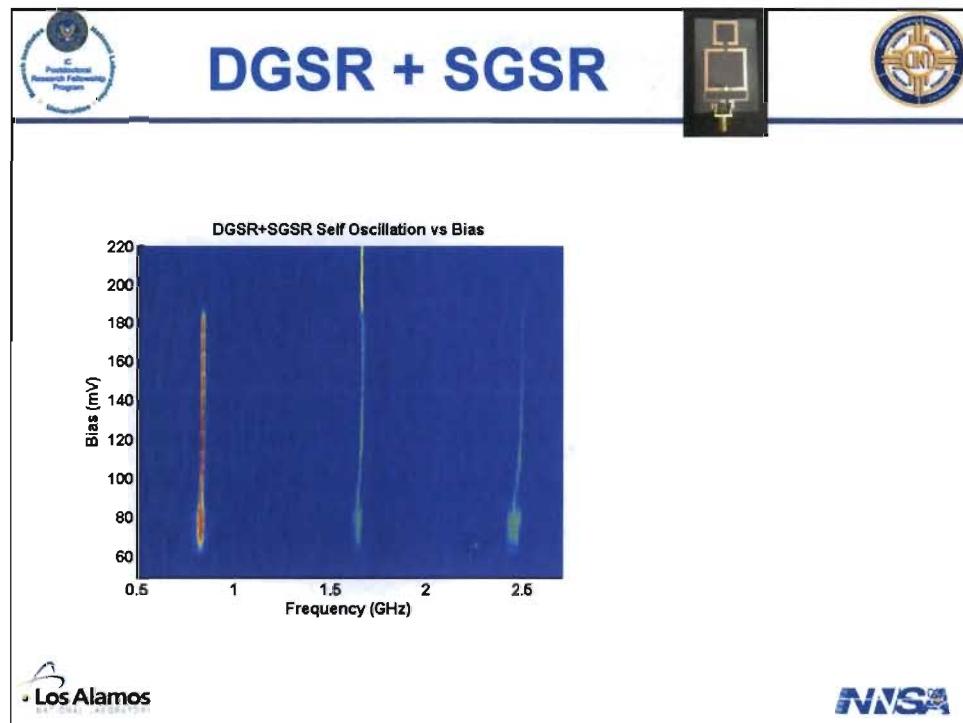
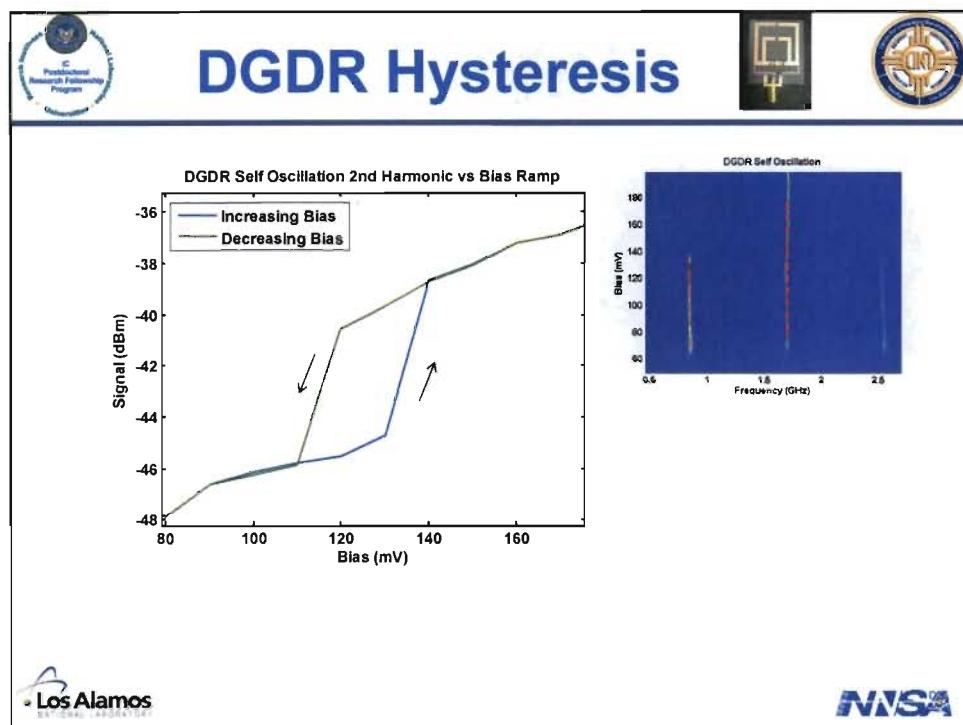
Frequency Locking

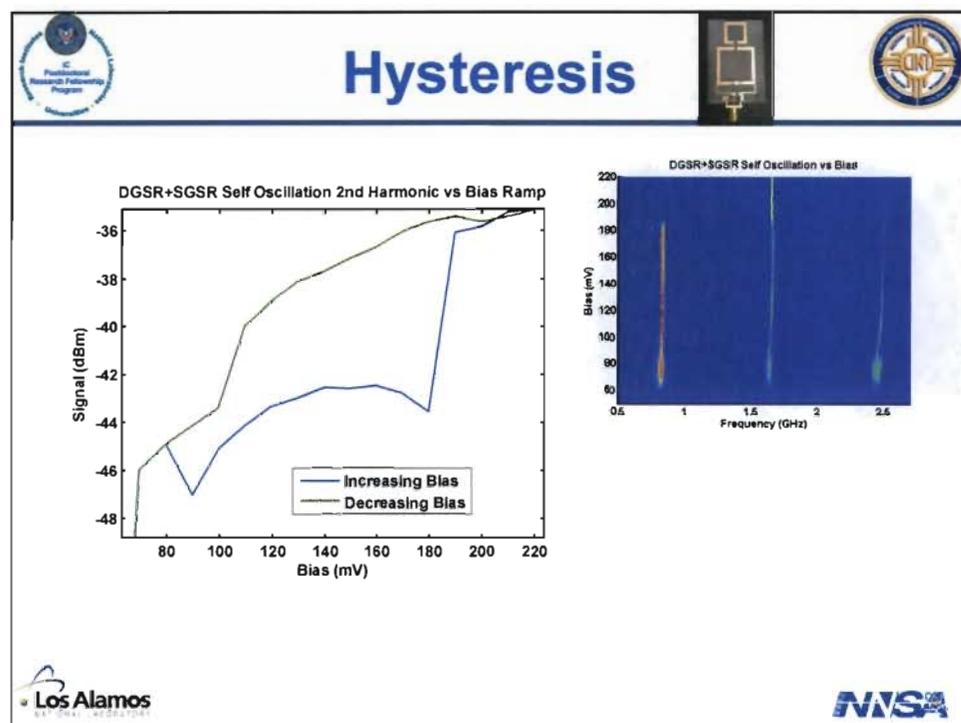
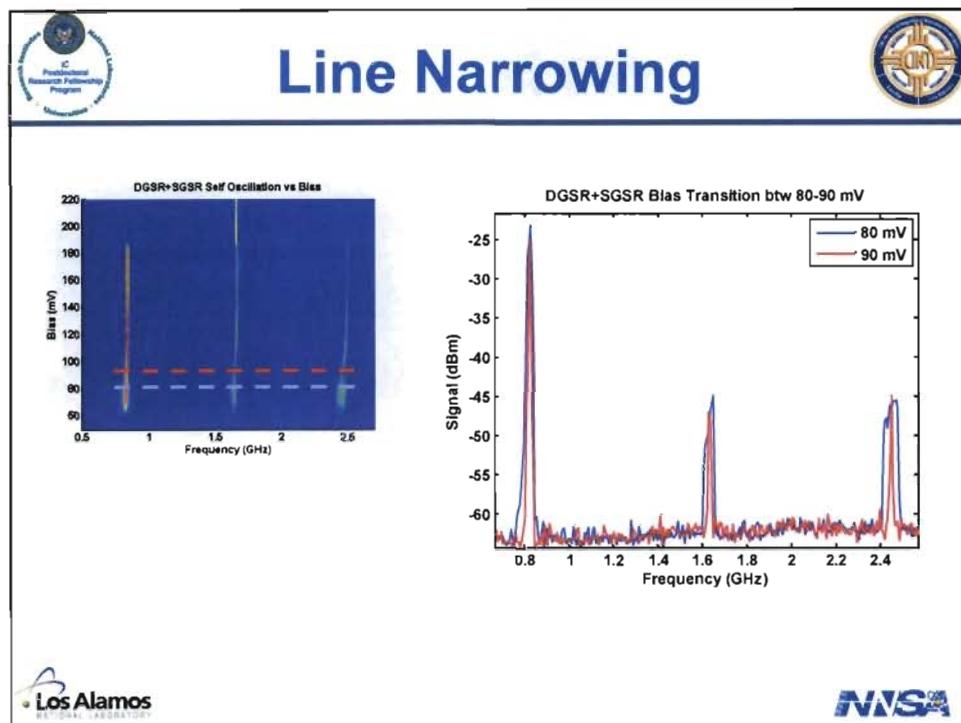
Frequency Locking w/ -30 dBm RF Input

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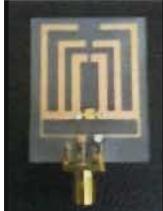






Double Gap Multi-Ring

No self oscillation observed...
at least below 2.9 GHz.



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